

# **MEMS Energy Harvesting & Conversion Devices enabled by patterned Super-magnets leading to Internet of Things (IoT)**

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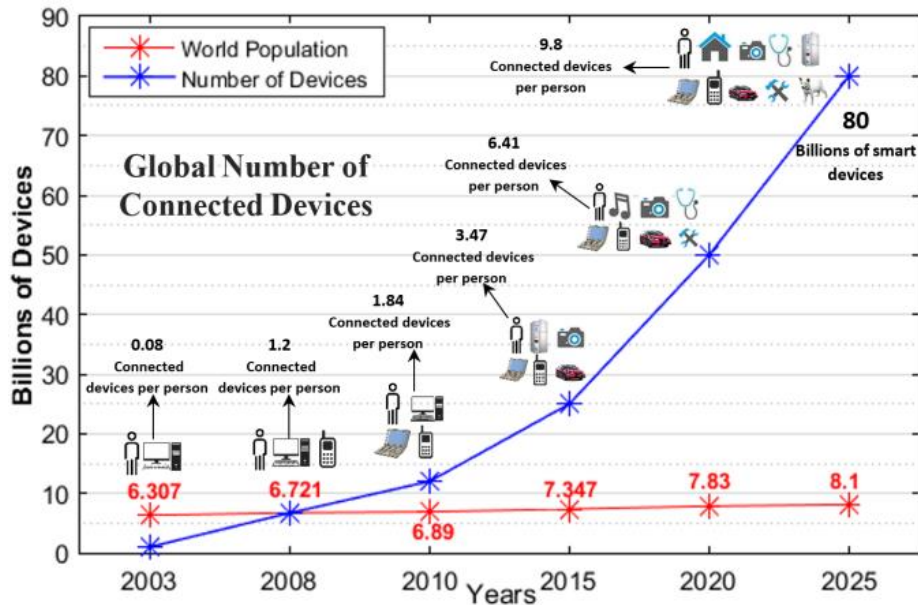
**EnerHarv 2022**  
**Raleigh, Durham, USA, 5-7 April 2022**

- ❖ **Powering IoT – Vibration (mechanical) Energy Harvesting (VEH)**
- ❖ **Electromagnetic (EM) Vibration Energy Harvesters**
  - Linear (narrowband) to Nonlinear (wideband)
- ❖ **Scaling Issues of EM-VEH Devices: MESO to MEMS scale**
  - Challenges and Roadmaps
- ❖ **Power conversion - Employing Micro-Nano-Magnetics**
- ❖ **Roadmap and Future Directions**

# Powering IoT - Vibration (mechanical) Energy Harvesting

# Internet of Things (IoT) & Energy Harvesting

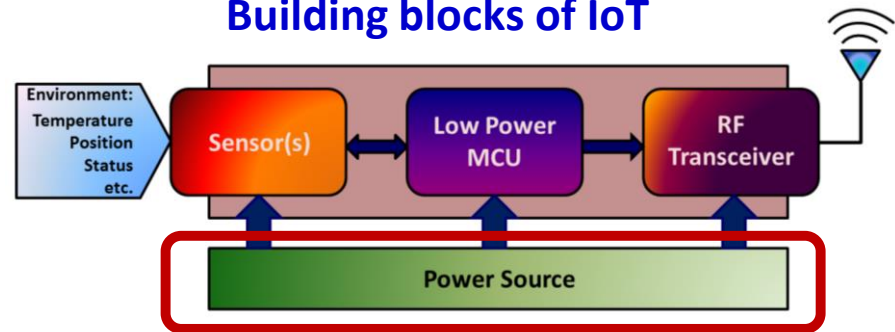
**Internet of Things (IoT)** – Wireless sensors/nodes connecting things / devices to internet



Farhan et al., Energy Efficiency for Green Internet of Things (IoT) Networks: A survey, Network 2021, 1, 279-314


## Wireless Sensor Nodes & associated Networks

### Building blocks of IoT




**Self-powering?**

**Batteries**




**Not suitable for  
'fit-and-forget'  
Applications!  
Costs? Pollutions?**


**Energy Harvesting**



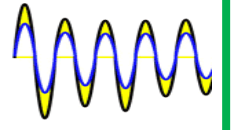
**Solar**



**Thermal**



**RF**

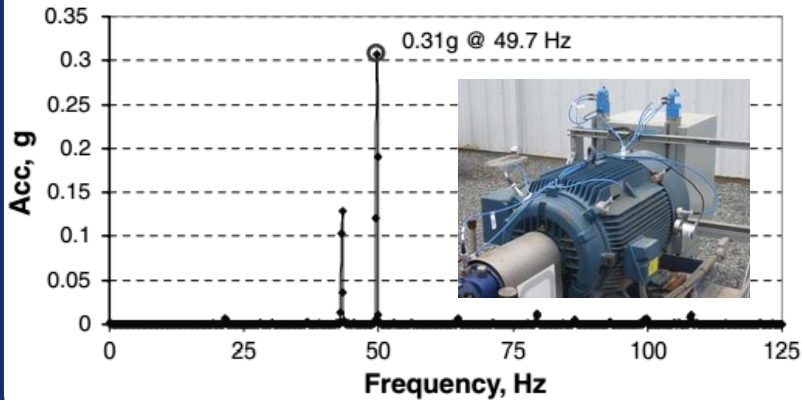


**Vibration**

# Vibration Energy Harvesting (VEH)

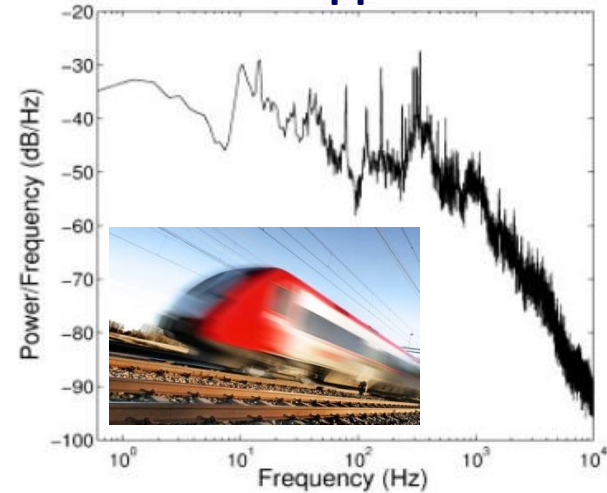
Mechanical energy is most easily accessible – **Ambient Vibrations**  
Low amplitude vibrations can be found in every sphere of life – **Huge Opportunity**

## Acoustic Emission Monitoring System



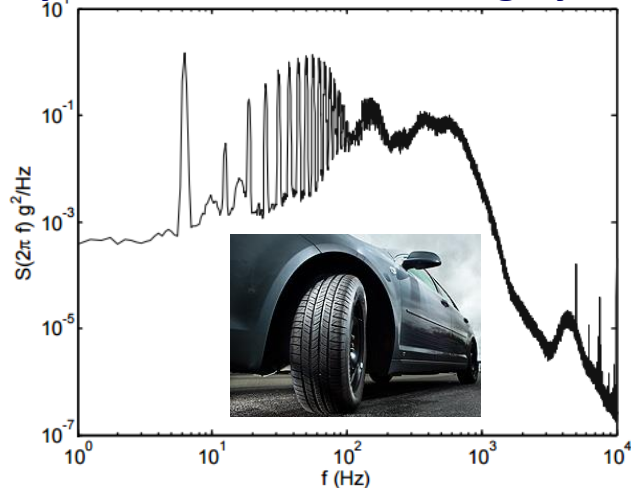
S. Beeby et al, J. Micromech. Microeng. 17, 1257–1265 (2007)

## Railcar Application



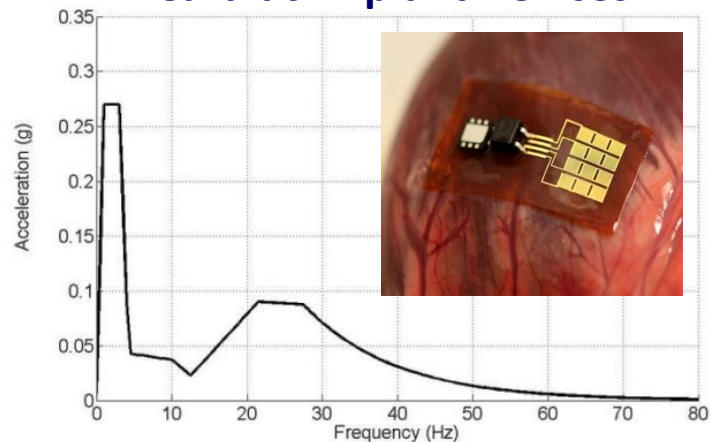
H. Vocca et al., Kinetic energy harvesting, in FAGAS G. (ed.), ICT Energy Concepts Towards Zero Power ICT, pp. 25-48 (2014)

## Tyre Pressure Monitoring System



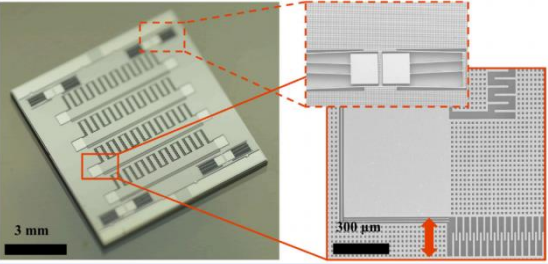
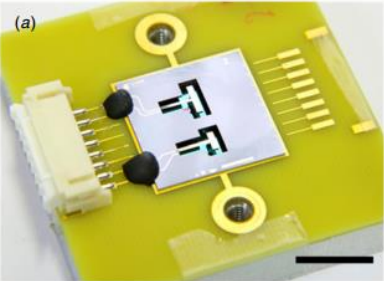
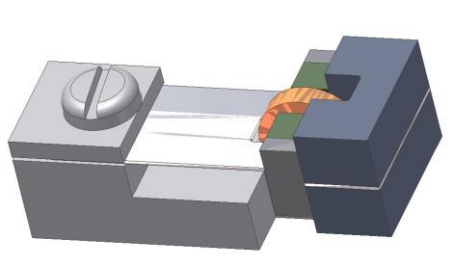
M. Löhndorf et al., Proc. PowerMEMS 2007, pp 331–4 (2007)

## Cardiac Implant Devices

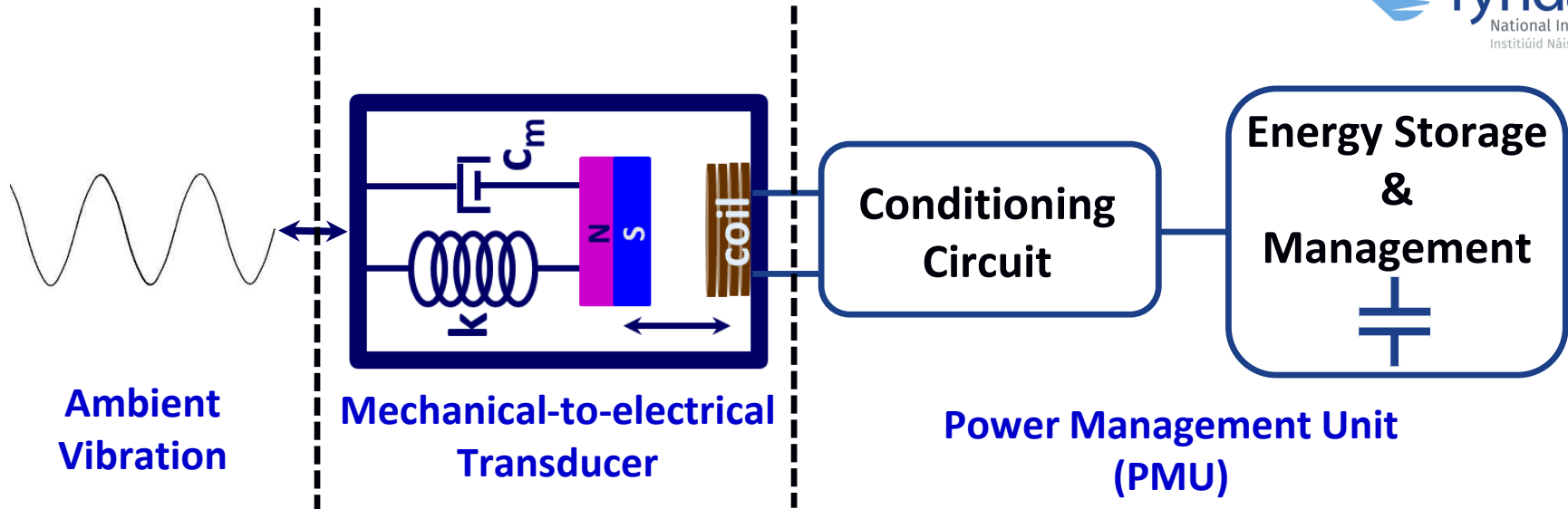


M. Deterre et al., Proc. Symp. DTIP MEMS/MOEMS, pp. 387–391 (2011)

## Comparison of Different Transduction Methods

Electrostatic	Piezoelectric	Electromagnetic
<ul style="list-style-type: none"><li>• Based on <b>changing capacitances</b> of the charged plates.</li><li>• The voltage will change as the capacitance changes.</li></ul>	<ul style="list-style-type: none"><li>• <b>Charge displacement</b> occurs when the material is strained.</li><li>• Thus a potential difference is obtained.</li></ul>	<ul style="list-style-type: none"><li>• Based on the <b>Faraday's Law of Induction</b>.</li><li>• When a conductor moves through a magnetic field, a potential difference is induced.</li></ul>
		
<ul style="list-style-type: none"><li>• Suitable for MEMS</li><li>• Low o/p voltage at high operational voltage</li><li>• High impedance values</li></ul>	<ul style="list-style-type: none"><li>• Active materials for fabrication, brittle.</li><li>• High Voltage, Low Current</li><li>• High Impedance values</li></ul>	<ul style="list-style-type: none"><li>• Do not need any extra component.</li><li>• High Current, Low Voltage</li><li>• Challenging in MEMS</li></ul>

# Inertial Energy Harvesting System



## Second Order Spring Mass Damper System

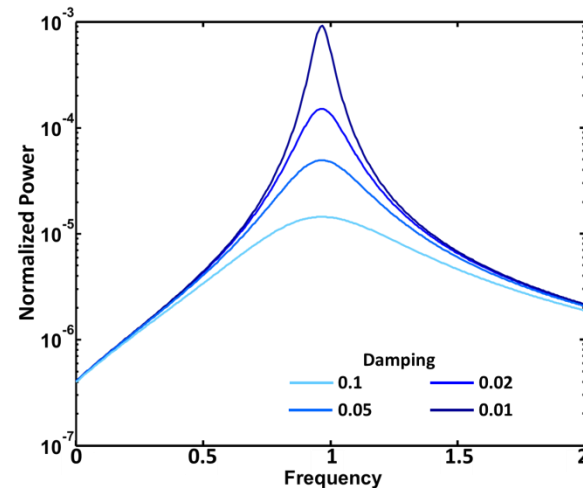
$$m\ddot{x}(t) + (c_m + c_e)\dot{x}(t) + kx(t) = -m\ddot{y}(t)$$

Power dissipated in the electrical damper

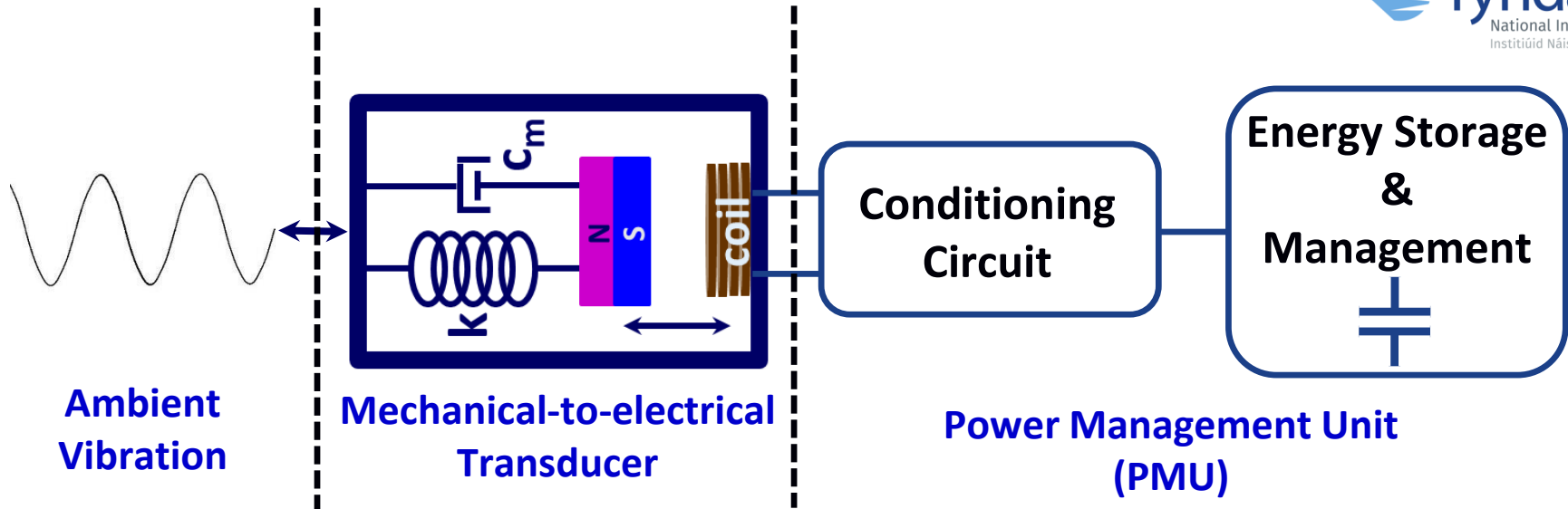
$$P_{\text{elec}} = \frac{c_e \left(\frac{\omega}{\omega_n}\right)^3 Y_0^2 \omega^3}{2\omega_n \left[ \left\{ 2\rho_T \left(\frac{\omega}{\omega_n}\right) \right\}^2 + \left\{ 1 - \left(\frac{\omega}{\omega_n}\right)^2 \right\}^2 \right]}$$

At Resonance:

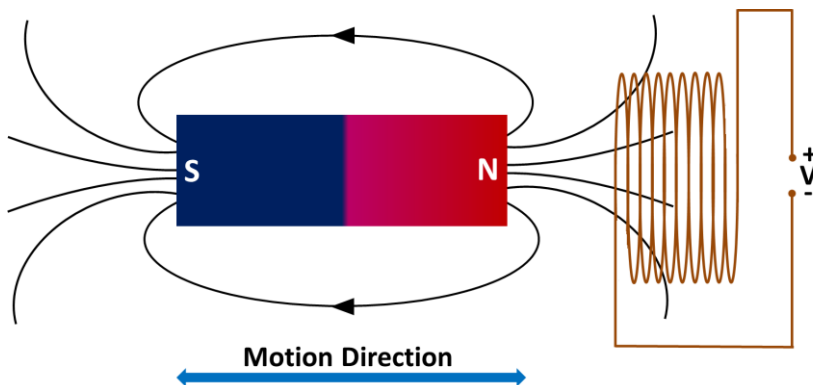
$$P_{\text{elec}}|_{\omega=\omega_n} = \frac{c_e m^2 Y_0^2 \omega_n^4}{2(c_m + c_e)^2}$$



# Inertial Energy Harvesting System



Principle governing EM generators is Faraday's Law



$$\text{EMF (V)} = -N \frac{d\Phi}{dt}$$

$$\text{Induced voltage } V_o = Z_{max} \omega_n N \frac{d\Phi}{dz}$$

$$\text{Load Voltage } V_L = V_o \frac{R_L}{(R_L + R_C)}$$

$$\text{Power delivered to load } P_L = \frac{V_L^2}{2R_L}$$

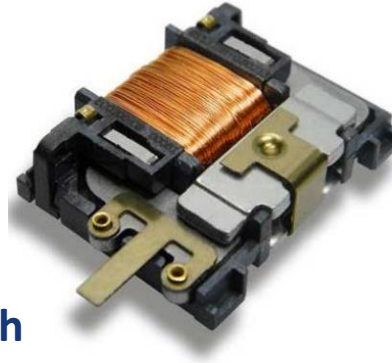


# **Electromagnetic (EM) VEH: Linear (narrowband) to Nonlinear (wideband)**

# Commercially Available EM VEHs

## EnOcean Light Switch

- Energy output 120-210  $\mu\text{J}$  at 2V
- Dimension: 29 X 19 X 7  $\text{mm}^3$
- Application: Wireless Light Switch



## Perpetuum Energy Harvester

- Bulky – weigh 1.03 kg
- 4.2 mW, 8V, 0.525 mA @0.05g
- Applications: Railroad, Industrial monitoring



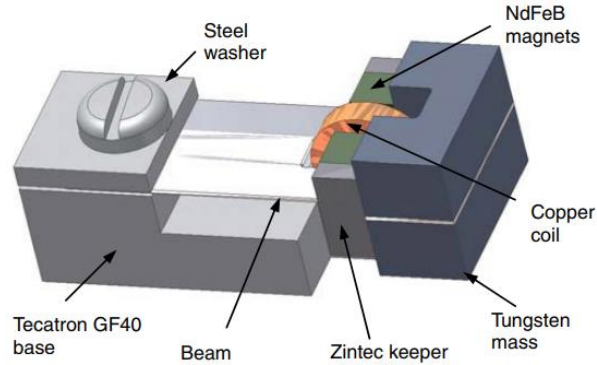
## ReVibe Energy Harvester

- modelD: 61 mm (height) x 32 mm (diameter)
- 2 mW @ 60 Hz and 0.1g
- Application: Industrial monitoring

## Remarks:

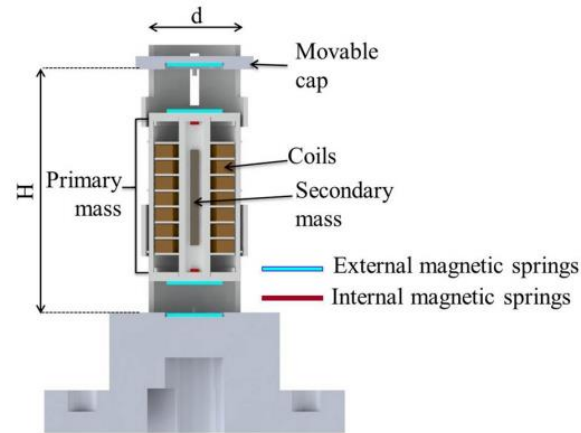
- Majority of them are bulky, heavy systems
- Most commercial products are linear or single frequency
- Focuses on structural, infrastructure and industrial monitoring applications
- Need for miniaturization is clear

# Macro/Mesoscale EM VEHs – SOA in literature



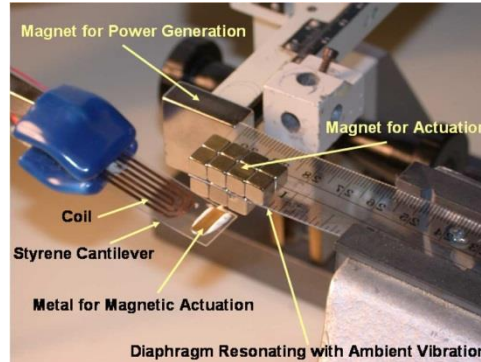
**S. P. Beeby et al., J. Micromech. Microeng., 17, 1257 (2007)**

- Volume – 0.15 cm<sup>3</sup>
- Power – 0.046 mW @ 0.06g
- Frequency – 52 Hz



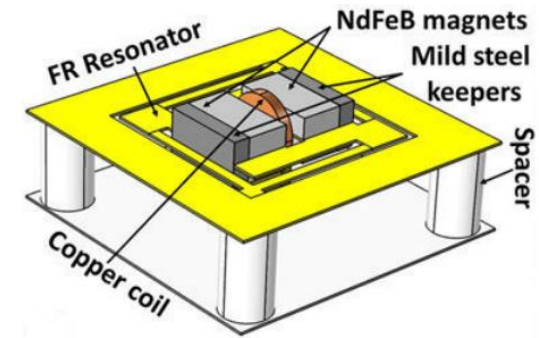
**V. Nico et al., Appl. Phys. Lett., 108, 013902 (2016)**

- Volume – 12.1 cm<sup>3</sup>
- Power – 2.75 mW @ 0.4g
- Frequency – 11.25 Hz



**H. Kulah, H. Najafi., IEEE Sens., 8(3), 261 (2008)**

- Volume – 0.3 cm<sup>3</sup>
- Power – 120 μW @ 1g
- Frequency – 64 Hz



**D. Mallick, S. Roy, Sens. Act. A, 226, 154 (2015)**

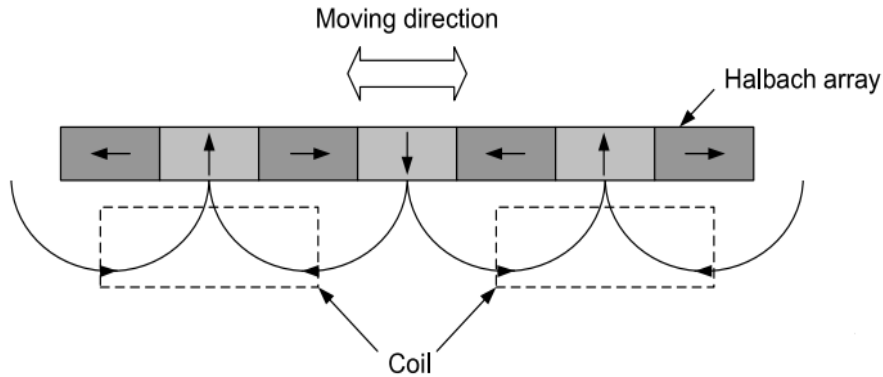
- Volume – 2.65 cm<sup>3</sup>
- Power – 0.5 mW @ 0.3g
- Frequency – 58.6 Hz

- ❖ Volume – 0.1 cm<sup>3</sup> to higher
- ❖ Discrete components assembled together – Fine machining
- ❖ Magnet array to increase flux density
- ❖ Lower Frequencies – Application relevant
- ❖ Reasonable Power Density

# Magnetic Flux Distribution

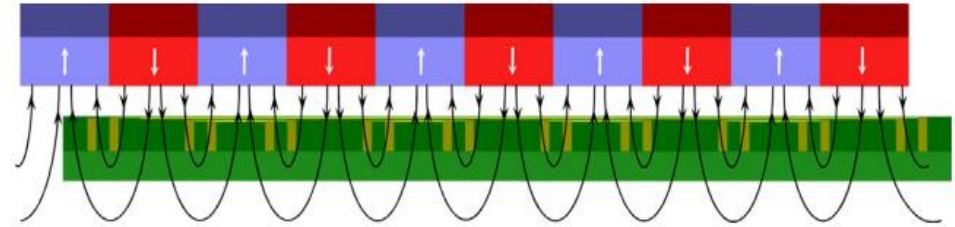
## Suitable Magnet Coil Arrangement – High Power Generation

### Halbach Array



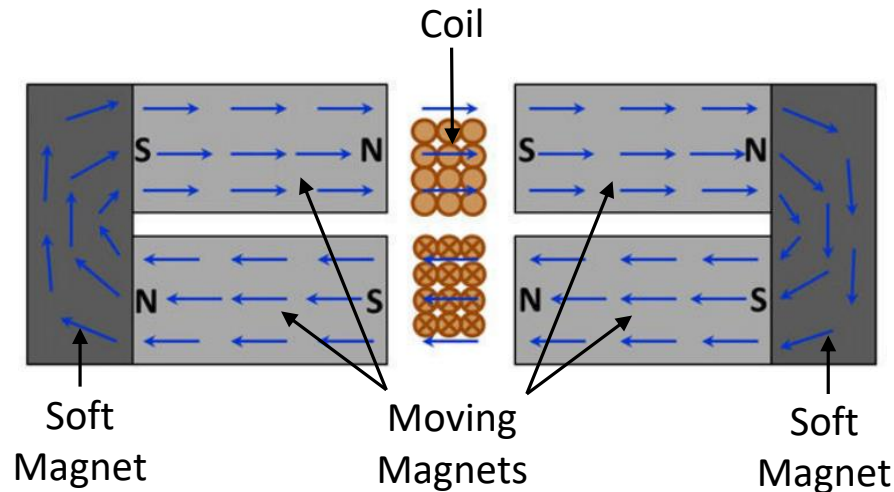
D. Zhu et al., *Smart Mater. Struc.*, 21(7), 075020 (2012)

### Alternate Oppositely Polarized Magnets



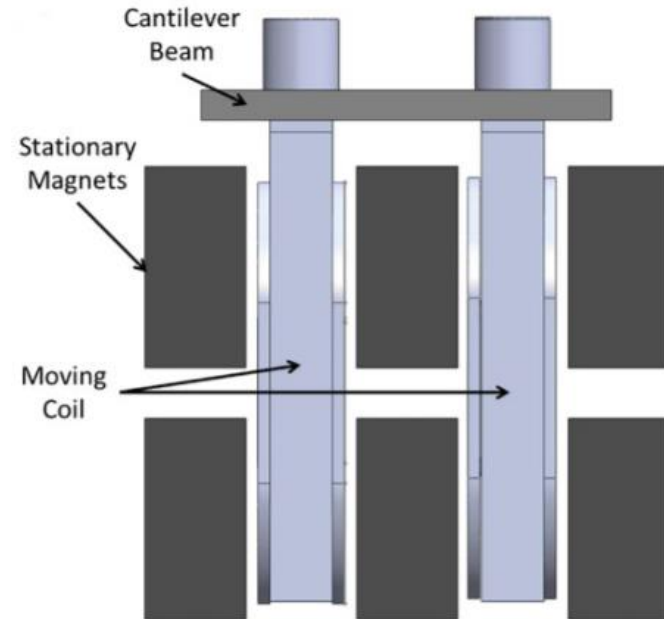
S. Roundy, E. Takahashi, *Sens. Act. A:Phys.*, 195, 98 (2013)

### Closed Magnetic Circuit



D. Mallick, S. Roy et al., *Smart Mater. Struc.*, 24(12), 122001 (2015)

### Double Cell Configuration

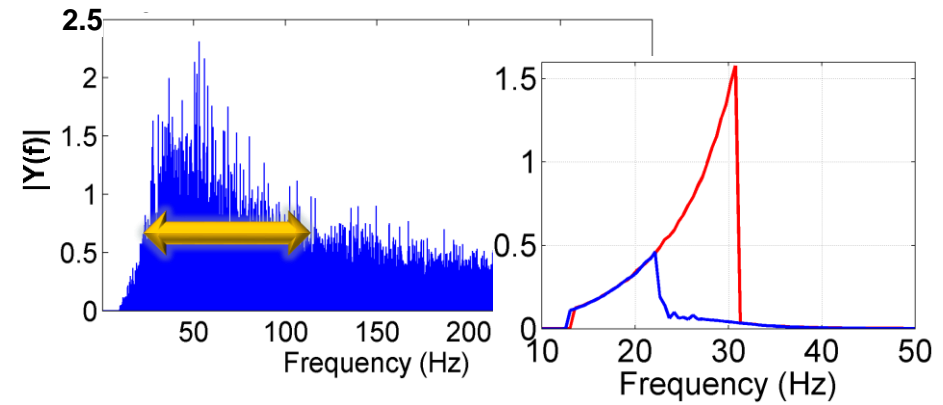
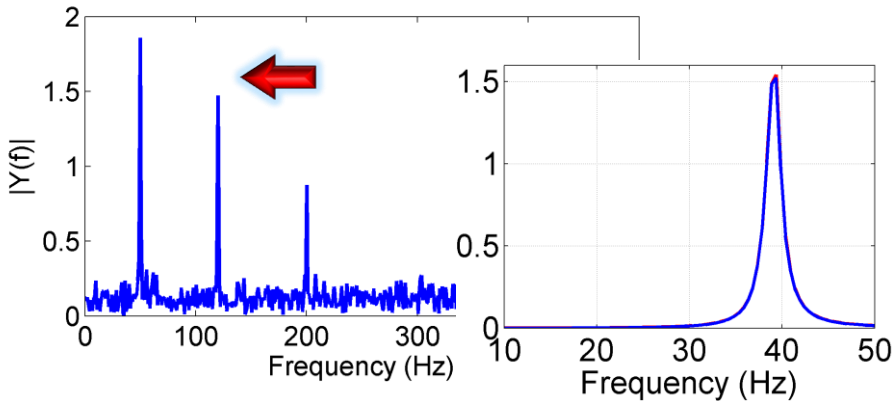


A. Marin et al., *J. Phys. D: Appl. Phys.* 44, 295501 (2011)

# Vibration Sources – low frequency

## Resonant, Impulse, Shock

## Broadband, Random, Noise



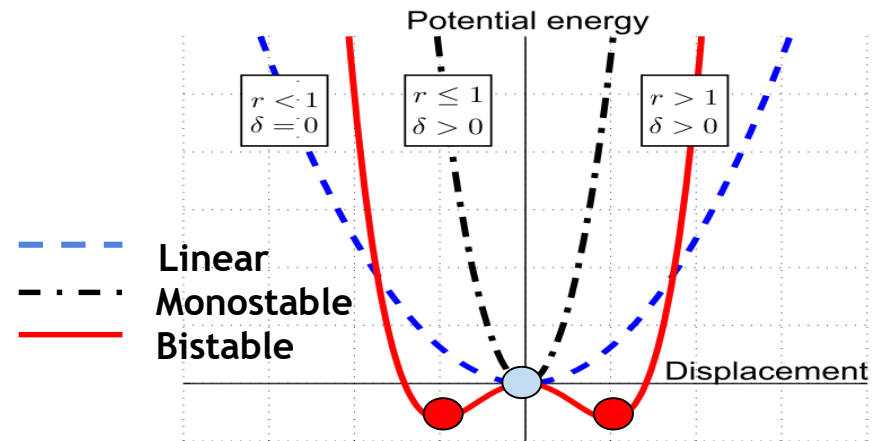
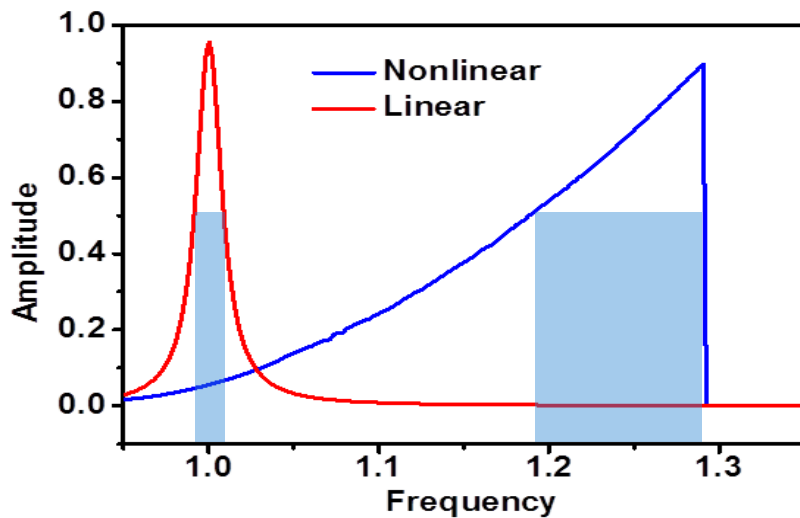
**Linear VEH**

**Wideband VEH**



# Nonlinear Energy Harvesting

Nonlinearity introduced through modified stiffness of the devices →  
**broader frequency response**



$$\ddot{x} = -\frac{dU(x)}{dx} - \gamma\dot{x} + f(t)$$

$$U(x) = -\frac{1}{2}\alpha(1-r)x^2 + \frac{1}{4}\beta x^4 \quad \delta = \frac{\beta}{\alpha}$$

$\alpha, \beta$  are constants

$r$  - determines nature of nonlinearity

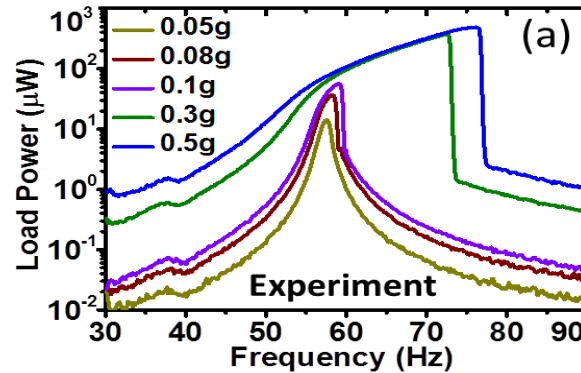
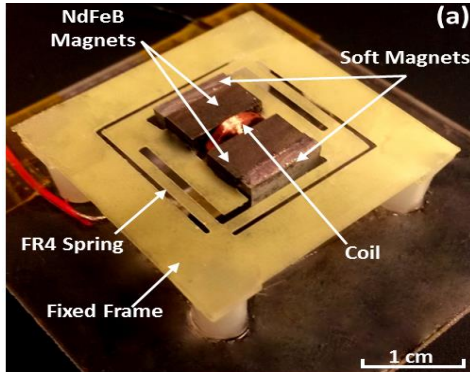
$$F(x) = -kx - k_n x^3$$





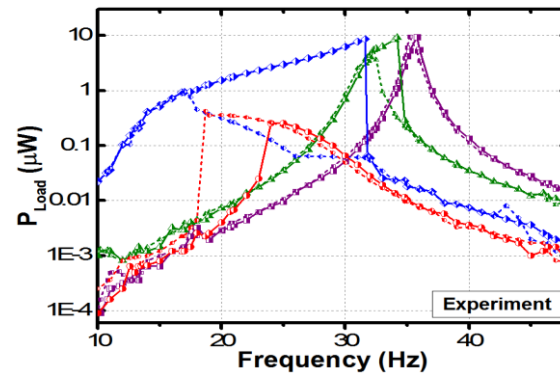
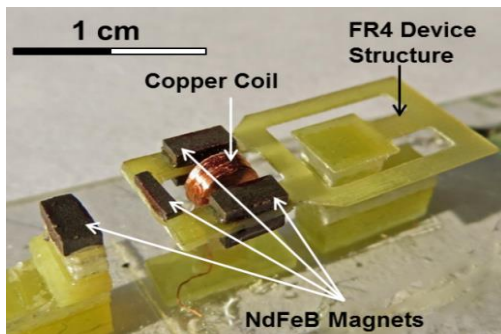
# Miniaturized Nonlinear EMPG Systems

## Mono-stable Nonlinear EMPG



- Mono-stable nonlinearity from stretching strain in addition to bending of fixed-guided beams.
- 0.5 mW of peak power under 0.5g acceleration

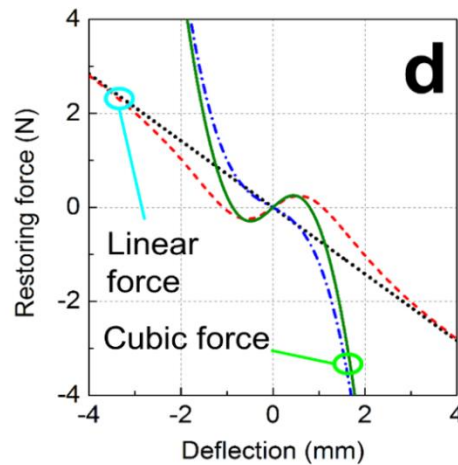
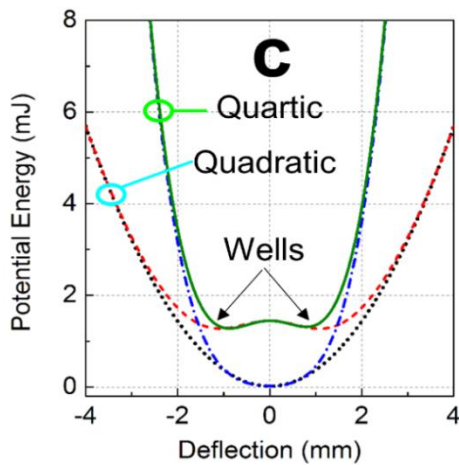
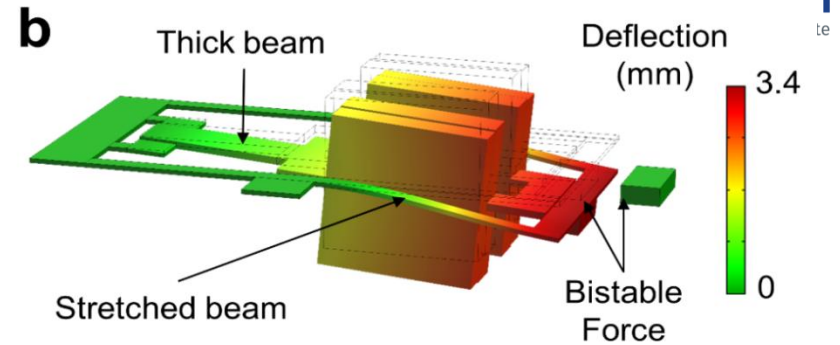
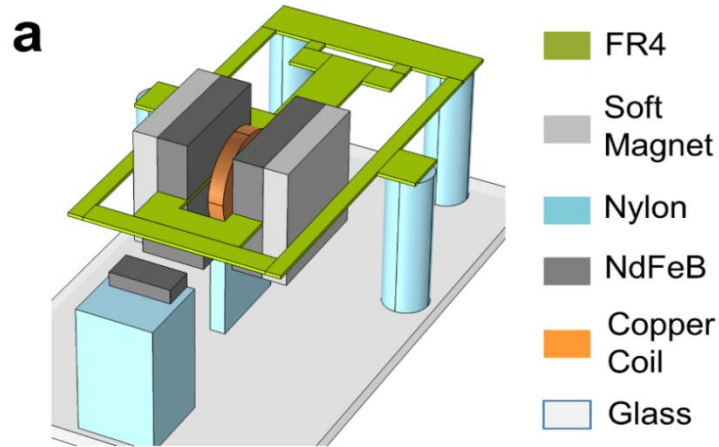
## Bi-stable Nonlinear EMPG



- Bi-stable nonlinearity from magnetic repulsive interaction.
- 29µW of power under 0.5g acceleration.

- ❖ D Mallick, A Amann, S Roy; Smart Materials and Structures 24, 015013, (2014)
- ❖ S Roy, P Podder, D Mallick; IEEE Magnetics Letters 7, 1-4, (2015)

# Combined bistable-quartic nonlinearity- Meso scale

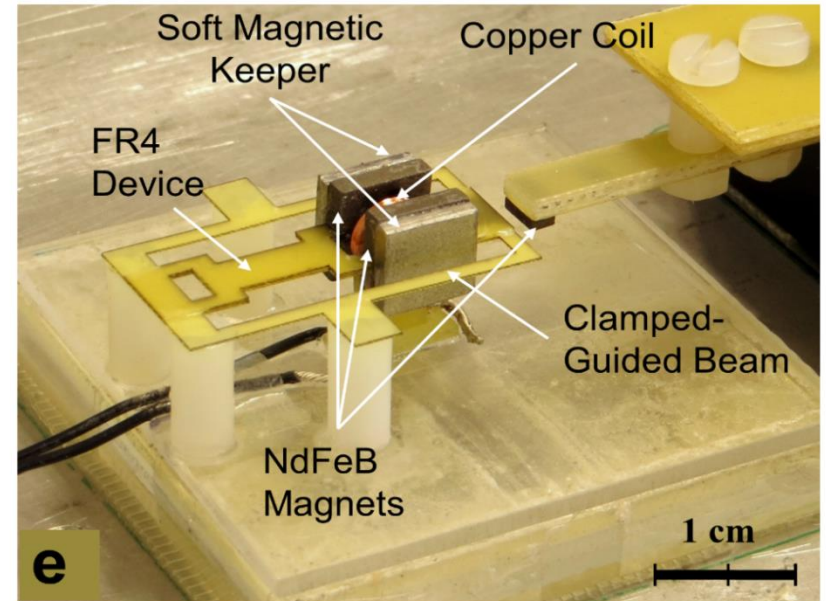


..... **Linear**

----- **BQD (Bistable-quadratic)**

- - - - **MQT (Monostable-quartic)**

— — — **BQT (Bistable-quartic)**



- ☐ Stretching of the **clamped-guided pair of beams** contributes to **mono-stable cubic nonlinearity**
- ☐ **Bi-stable nonlinearity** is incorporated by the **repulsive magnetic force** at the tip of the cantilever
- ☐ **1403  $\mu\text{W}$**  of power at 1g

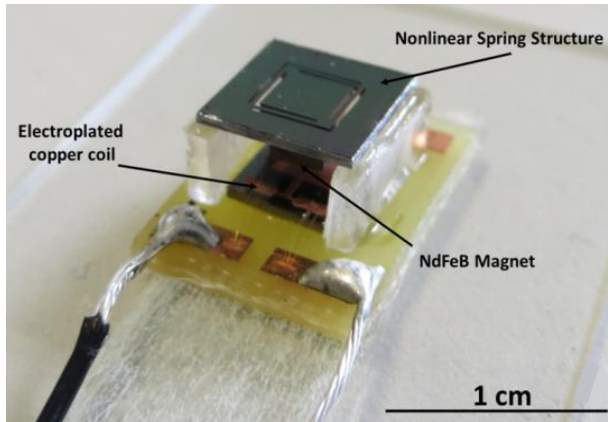
❖ P Podder, D Mallick, A Amann, S Roy; **Scientific Reports**; **6**, 37292, (2016)

❖ S Roy, P Podder, D Mallick, A Amman; **Patent granted – EP 3311472, Jan 2020; US 10971986B2, April 2021**



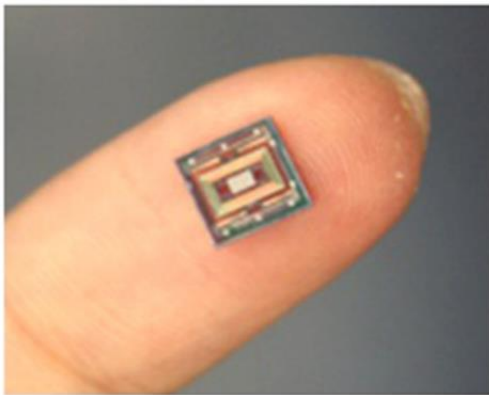
# Electromagnetic (EM) VEH: Meso to MEMS Scale

# MEMS-scale EM VEHs – SOA in literature



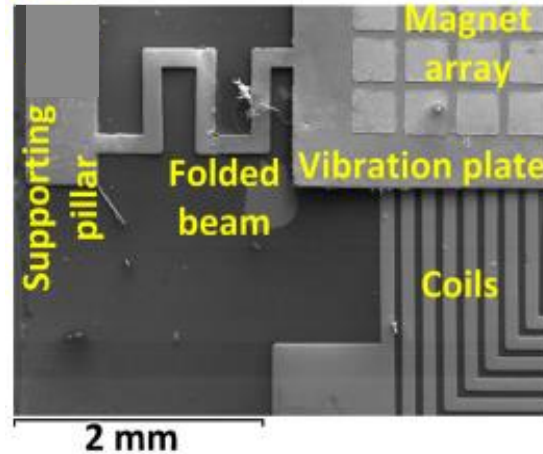
D. Mallick et al., *J. Microelectromech. Syst.*, 26(1), 273 (2016)

- Volume –  $0.1 \text{ cm}^3$
- Power –  $2.8 \mu\text{W}$  @  $0.5\text{g}$
- Frequency –  $630 \text{ Hz}$



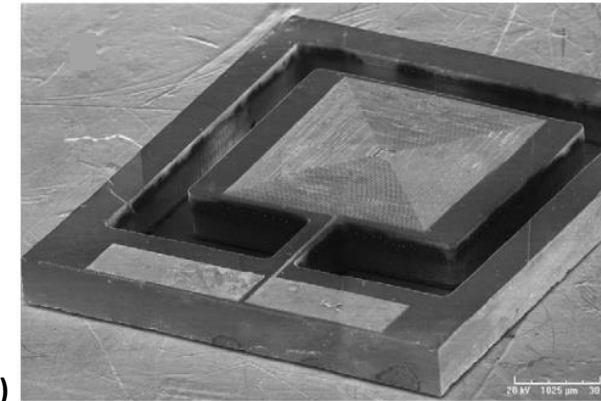
H. Liu et al., *Appl. Phys. Lett.*, 104, 053901 (2014)

- Volume –  $0.16 \text{ cm}^3$
- Power –  $0.06 \text{ nW}$  @  $1\text{g}$
- Frequency –  $384 \text{ Hz}$



M. Han et al., *Sens. Act. A: Phys.*, 219, 38 (2014)

- Volume –  $67.5 \text{ mm}^3$
- Power –  $10 \text{ nW}$  @  $1\text{g}$
- Frequency –  $48 \text{ Hz}$



S. Kulkarni et al., *Sens. Act. A: Phys.*, 145, 336 (2008)

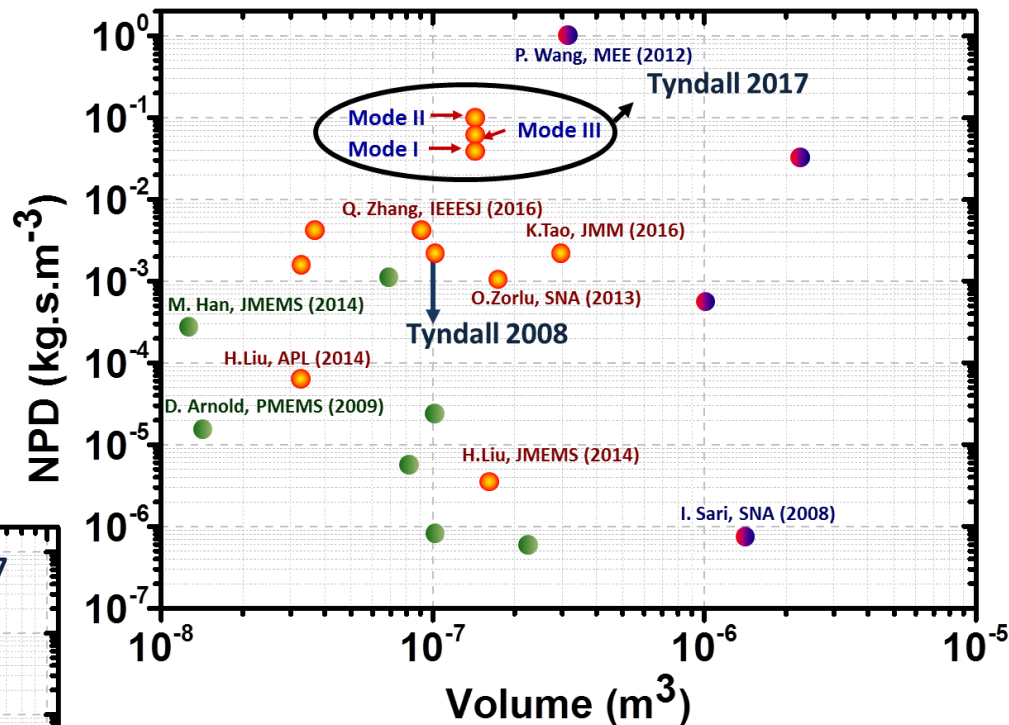
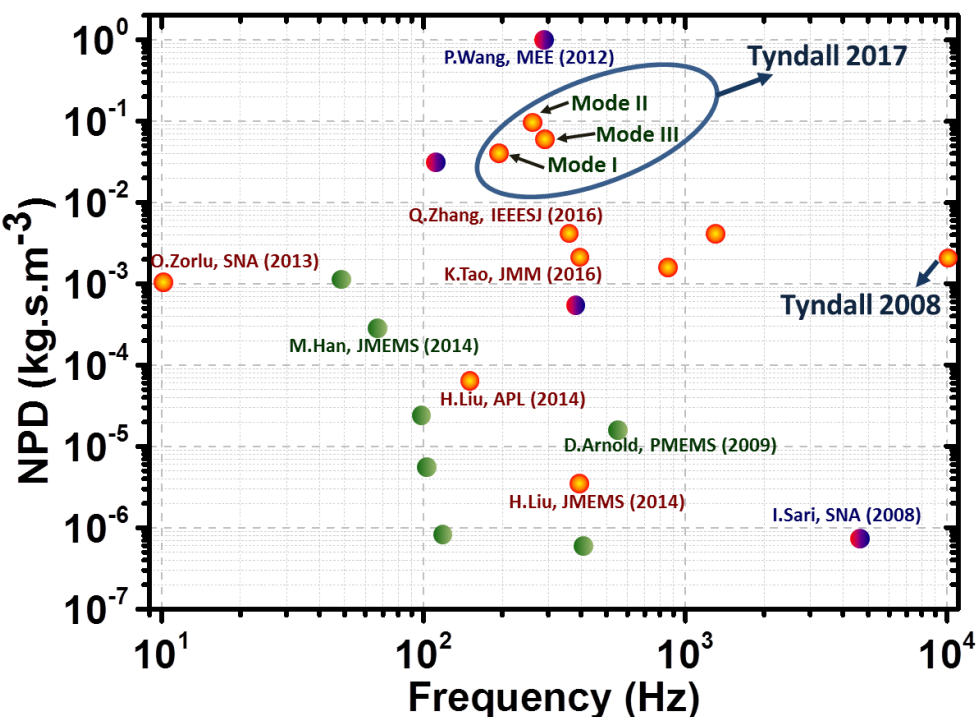
- Volume –  $0.1 \text{ cm}^3$
- Power –  $23 \text{ nW}$  @  $1\text{g}$
- Frequency –  $9830 \text{ Hz}$




- ❖ MEMS – micro-scale features
- ❖ Semi (most) or Fully (few) batch-fabricated
- ❖ Si or non-Si (PDMS, electroplated Ni) suspension systems
- ❖ Bulk Magnets

# Benchmarking the MEMS EM VEHs

Normalized Power Density(NPD)  

$$= \frac{\text{Power}}{\text{Acceleration}^2 \cdot \text{Volume}}$$

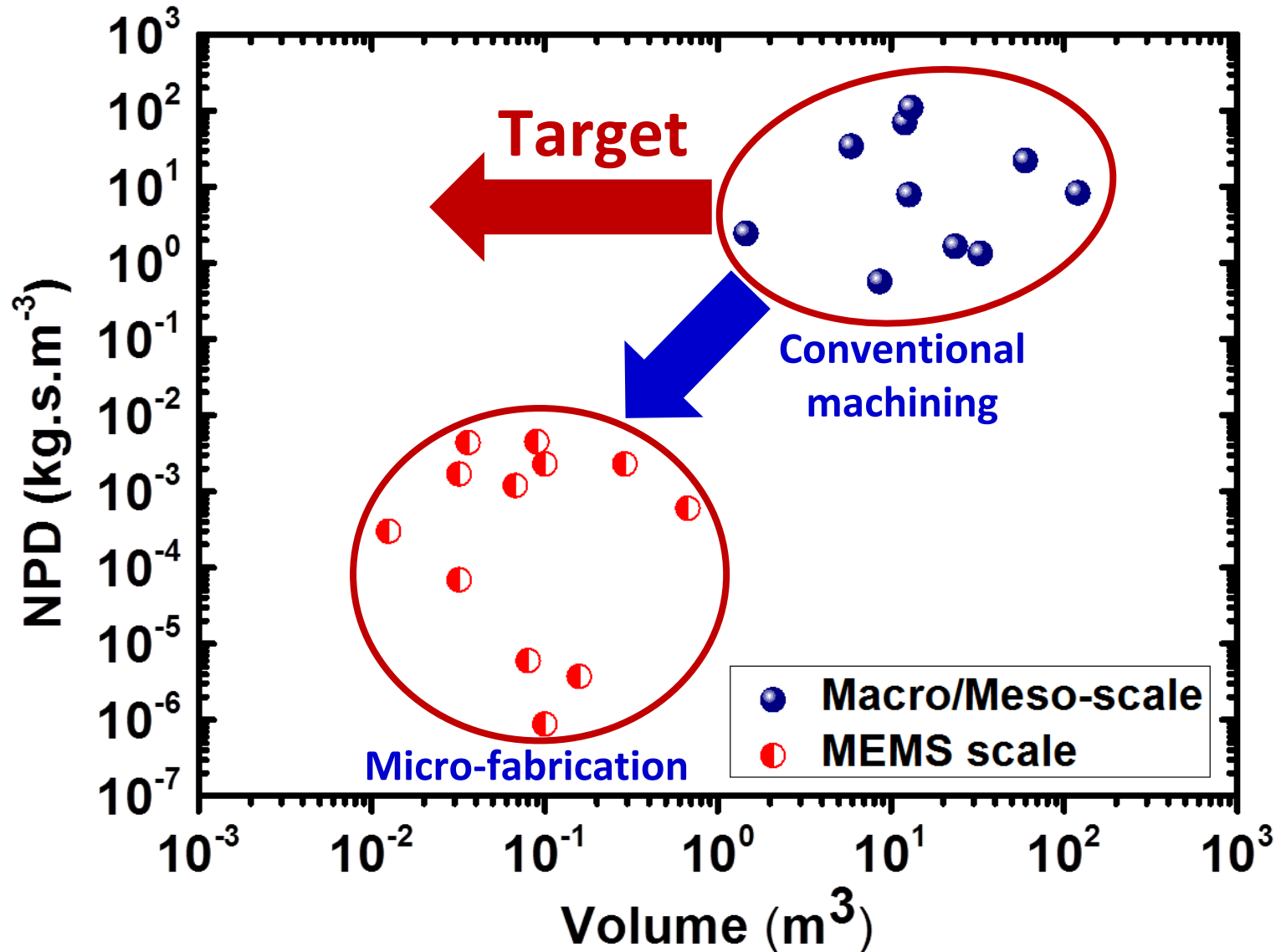


-  MEMS Suspension only
-  MEMS Suspension with micro-coil
-  Fully MEMS integrated

# Scaling remains the main issue

MEMS reduces cost

Trade Off: Size vs Effectiveness



# Scaling Issues of EM VEH Devices: Challenges and Roadmaps

# Scaling of EM VEHS

Maximum load power generated by a EM Generator at resonance for matched load

$$P_L = \frac{m^2 a_0^2}{8c_m \left( \frac{R_C c_m}{\gamma^2} + 1 \right)}$$

Where,  $\gamma = N \frac{d\phi}{dx}$  is the electromagnetic coupling

$$\begin{aligned} m &\propto L^3 \\ c_m &\propto L \\ \gamma &\propto L \\ R_C &\propto L^{-1} \\ \gamma^2 / R_C &\propto L^3 \end{aligned}$$



$$P_L \propto \frac{L^7 a_0^2}{L^2 + 1}$$

Two asymptotic scaling laws can be obtained:

$$P_L \propto L^5 a_0^2 \text{ For } L \rightarrow \infty$$

$$P_L \propto L^7 a_0^2 \text{ For } L \rightarrow 0$$

**Case I:**  $c_m \ll \frac{\gamma^2}{R_C}$  For  $R_C \rightarrow 0$

Coil damping dominates at large scale

$$P_L \approx \frac{m^2 a_0^2}{8c_m} \propto L^5 a_0^2$$

**Case II:**  $c_m \gg \frac{\gamma^2}{R_C}$

mechanical damping dominates at small scale

$$P_L \approx \frac{\gamma^2 m^2 a_0^2}{8c_m^2 R_C} \propto L^7 a_0^2$$

# Scaling of EM VEHs (II)

Let us take a look at the electromagnetic damping term

## Number of Coil Turns

- Decreases with scaling
- Limited by technology

$$c_e = \frac{N^2 \left(\frac{d\phi}{dx}\right)^2}{R_C + R_L}$$

## Magnetic Flux Linkage Gradient

- ❖ More important than flux density
- ❖ Magnetic flux density from PM scales independently of size
- ❖ Flux linkage gradient depends on area of the coil

## Coil Resistance

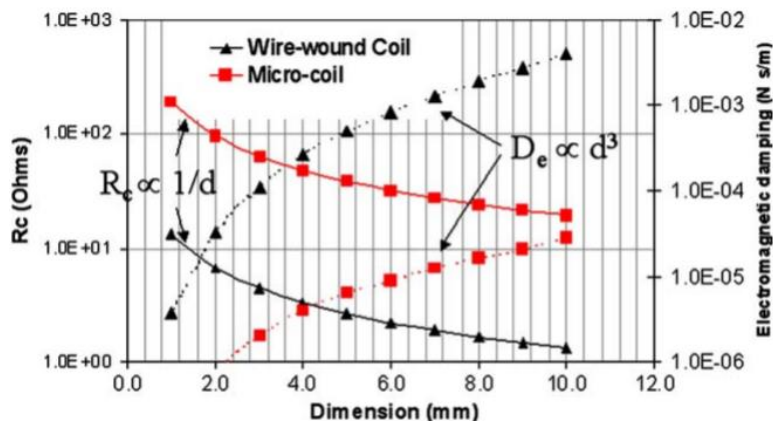
Micro-fabricated Coil Resistance:

$$R_C = \frac{8\rho_{Cu}(d_o + d_i)}{(d_o - d_i)^2} (4N^3 - 4N^2 + N)$$

$R_C$  is inversely proportional to the scaling factor

If generator scales by factor  $b$

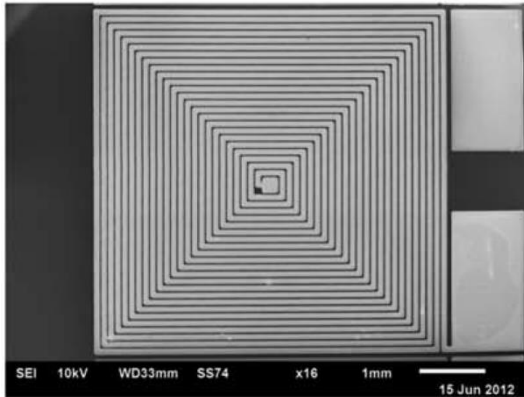
- Flux density gradient increases by  $b$
- Area of the coil decrease by factor  $b^2$
- Flux linkage gradient decrease by  $b$





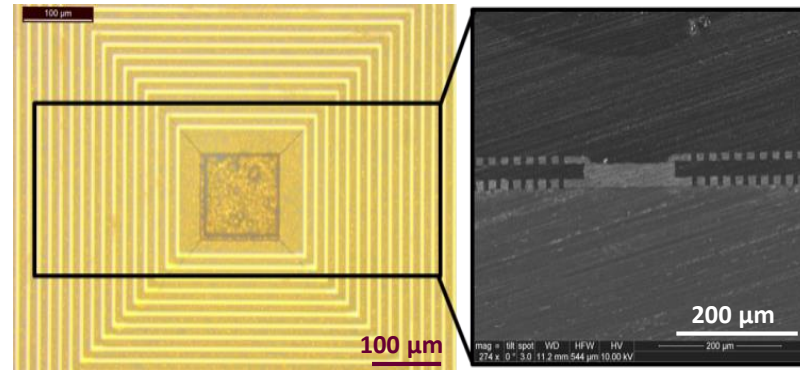
# Micro-fabricated Coils in EM-VEH

## Single/Multi-layer micro-coils



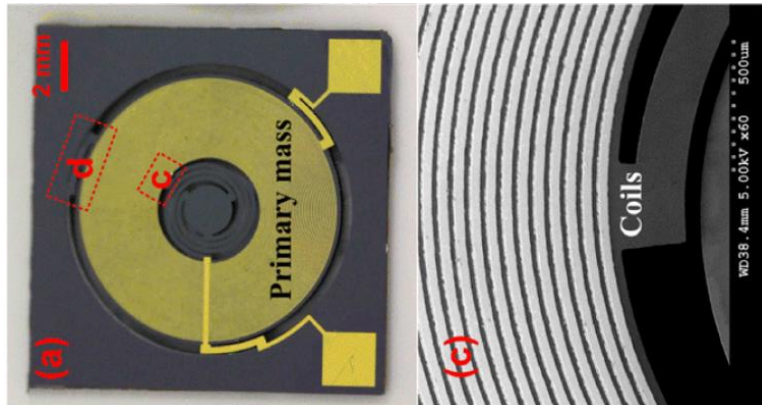
### Electroplated Single Layer Copper Coil<sup>1</sup>

- 20 Turns,
- 3.6  $\Omega$ ,
- 22  $\mu\text{m}$  thickness



### Electroplated Double Layer Copper Coil<sup>2</sup>

- 144 Turns,
- 190  $\Omega$ ,
- 10  $\mu\text{m}$  track width, spacing, thickness



### Sputtered Double Layer Cr/Au Coil<sup>3</sup>

- 57 Turns,
- 12.7 k $\Omega$ ,
- 30/30width/spacing, 350 nm thickness

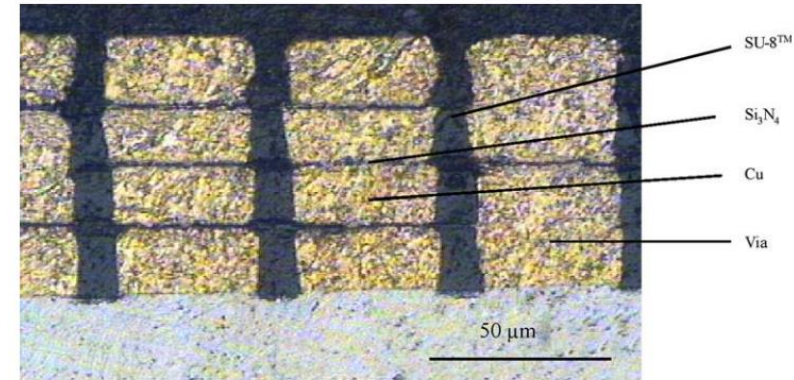
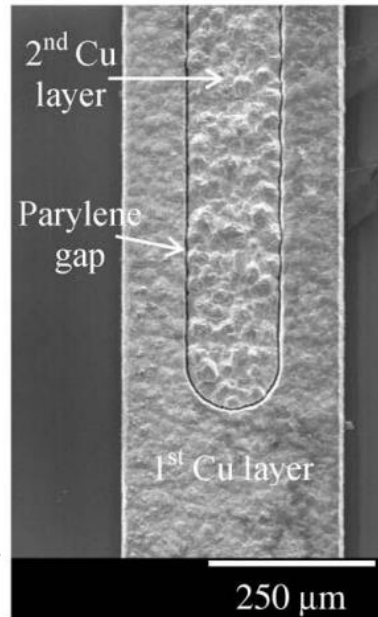
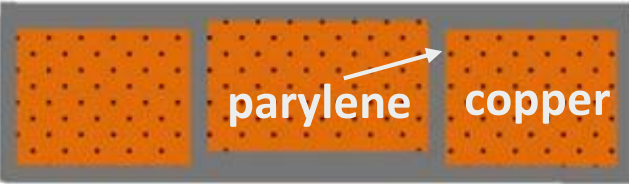
- ❖ Sputtered Al, Cr/Au coils – very thin, high resistance
- ❖ Electrodeposited Cu coils – thick, low resistance, low cost

1. Q. Zhang, E. S. Kim, Proc. IEEE, 102(11), 1747 (2014)
2. D. Mallick et al., J. Microelectromech. Syst., 26(1), 273 (2016)
3. K. Tao et al., J. Micromech. Microeng., 26, 035020 (2016)



# Roadmap for Micro-fabricated Coils

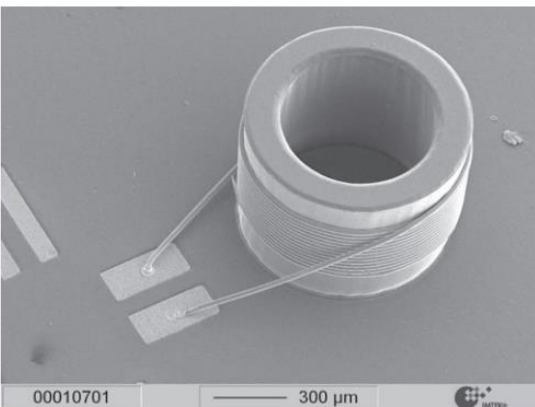
## Improving the packing density:



- Parylene conformally coated high density coil<sup>1</sup>
- Doubles the number of turns
- Complex fabrication process
- **Concern:** dielectric breaking - electrical short circuit

- Increasing the stack height<sup>2</sup>
- Increasing the aspect ratio (AR=w/h)
- Using standard lithography, it is possible to develop as high AR as 1:17 and 5 μm resolution<sup>3</sup>

## MEMS compatible 3D coil<sup>4</sup>:

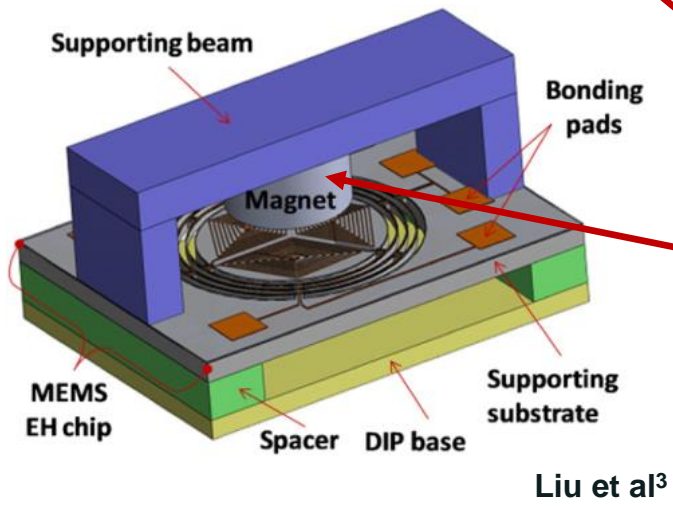
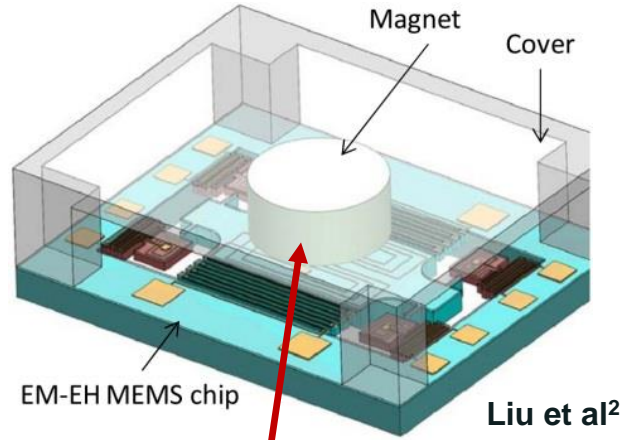
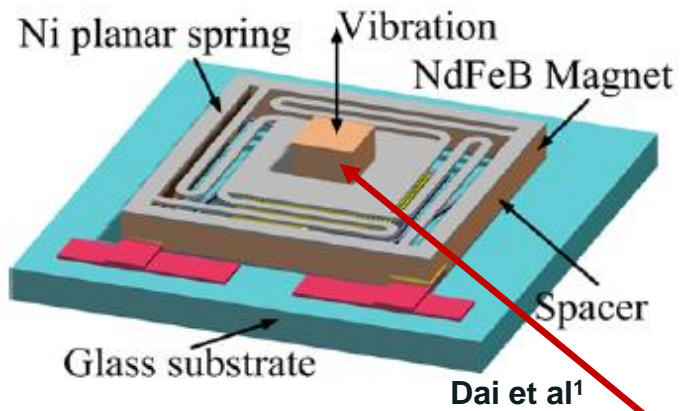


- Wire-bonded
- 3D solenoid structure
- Number of windings is limited by the post height

1. F. Herrault et al., *J. Microelectromech. Syst.*, 19(6), 1277 (2010)
2. C. Ruffert, H. H. Gatzert, *Microsyst. Technol.*, 14, 1589 (2008)
3. R. Anthony et al., *J. Micromech. Microeng.*, 26, 105012 (2016)
4. K. Kratt et al., *J. Micromech. Microeng.*, 20, 015021 (2010)

# Integrated Permanent Magnets

## Key Challenge: Permanent Magnet Integration

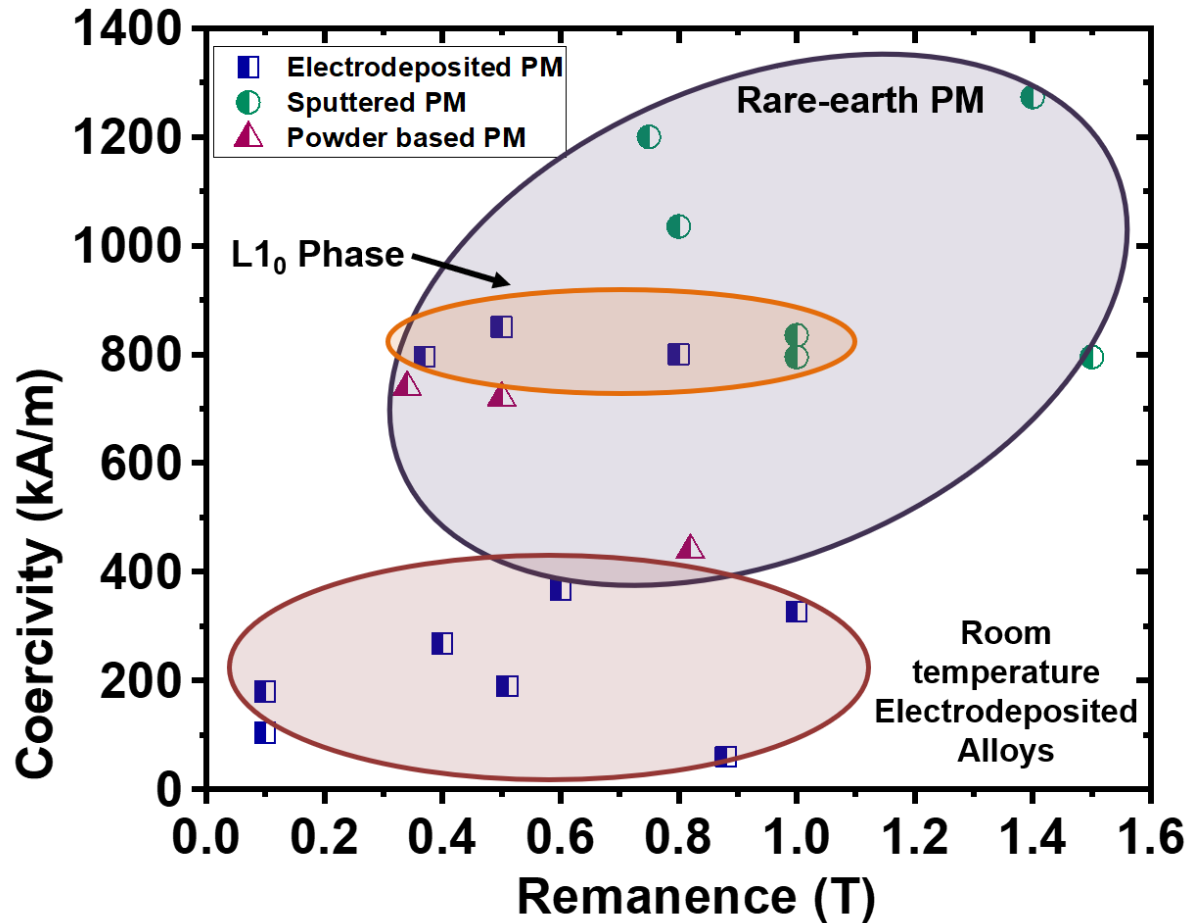


**Bulk NdFeB Magnet**

- Main meso-scale element in MEMS EM-VEHs
- Barrier in batch fabrication

1. X. Dai et al., Appl. Phys. Lett., 100:031902 (2012).
2. H. Liu et al., IEEE J. Microelectromech. Syst. 23(3): 740-49 (2014).
3. H. Liu et al., J. Micromech. Microeng. 22:125020 (2012).

## State-of-the-art Permanent Magnet Materials



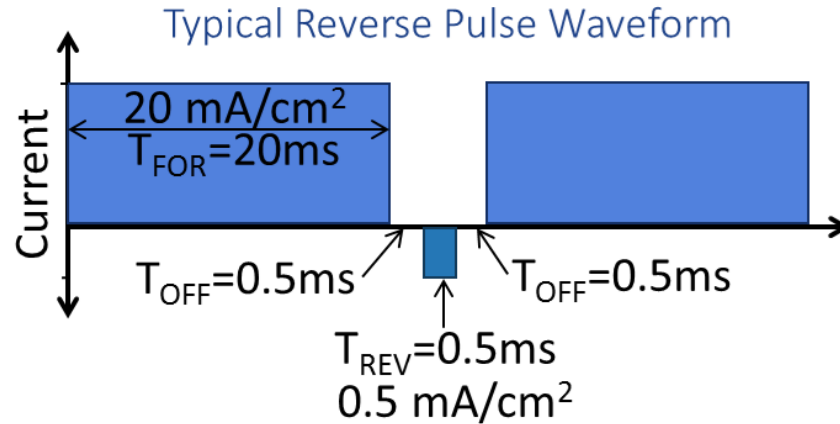
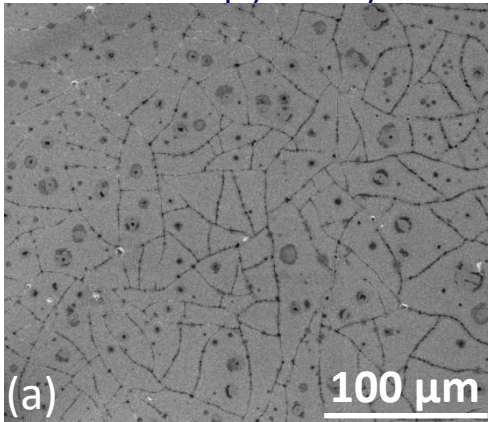
- **Rare Earth Magnets** – High Temperature processing – as such Not MEMS compatible
- **CoPt/FePt – L<sub>10</sub> phase** – CMOS compatible deposition – High temperature (700°C) annealing
- **CoPtP/CoNiMnP/CoNiReP - Co hcp phase** – **can become hard magnetic at room temperature**

# Integrated Permanent Magnet Development

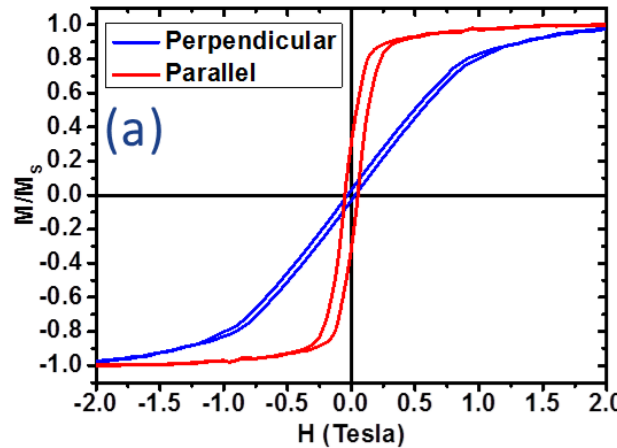
Electrodeposition – low cost, high deposition rate

Pulse Reverse Plating (PRP) of CoPtP Permanent Magnets – Low Stress

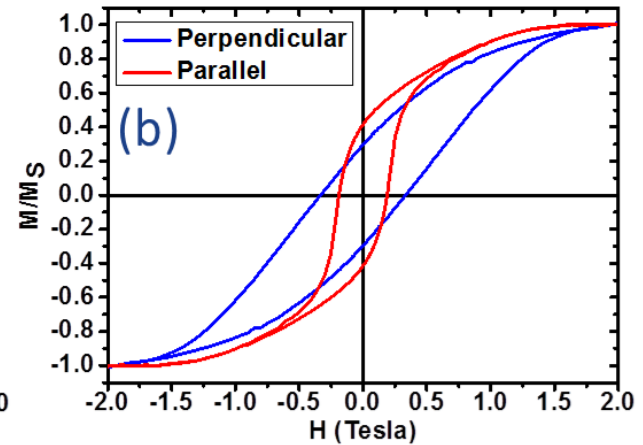
Conventional DC Plated Film  
Room Temp., 20 mA/cm<sup>2</sup>



SQUID Magnetic Measurement Hysteresis Loop

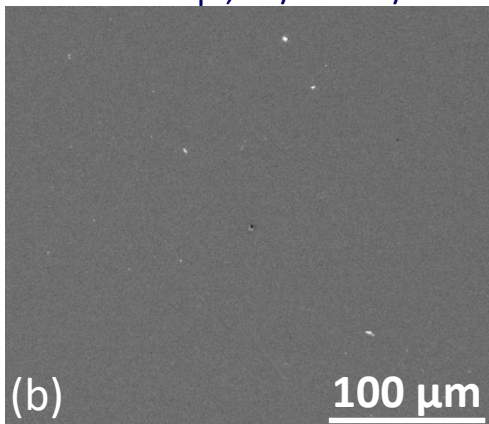


$H_C (\parallel) - 24 \text{ kA/m}$   
 $H_C (\perp) - 41.4 \text{ kA/m}$



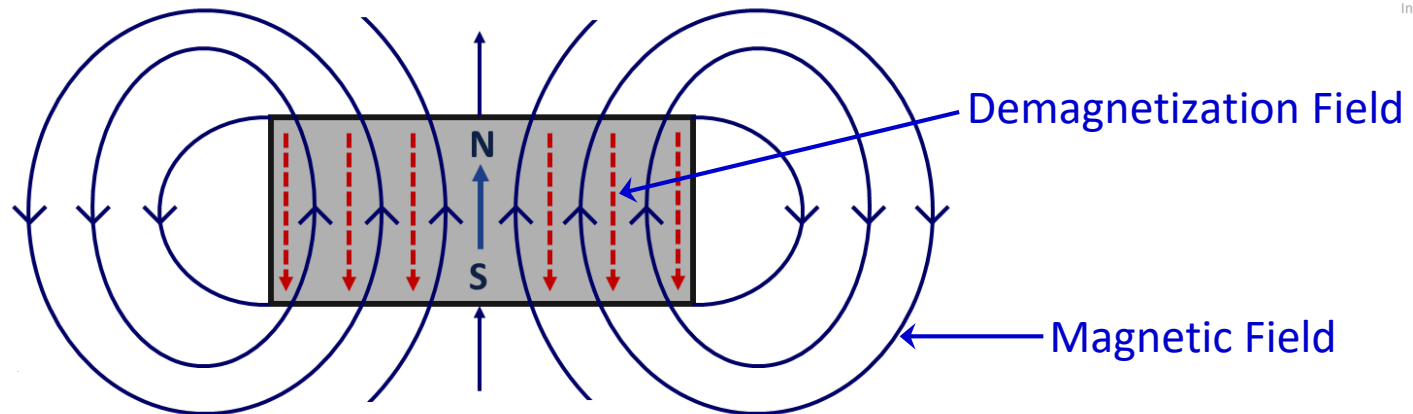
$H_C (\parallel) - 175 \text{ kA/m}$   
 $H_C (\perp) - 268 \text{ kA/m}$

Optimized PRP Plated Film  
Room Temp., 20/10 mA/cm<sup>2</sup>



The grain size of the films increase from 6.4 nm to 22.5 nm as the thickness grows from 0.9 μm to 26 μm

# Design Challenge – Integration of PM



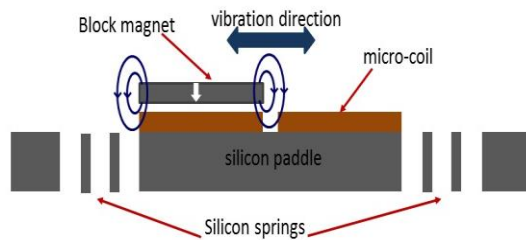
- The internal field  $H$  acting in a sample  $H = H_{ext} + H_d$   
 $H_{ext}$  - External field  
 $H_d$  - Demagnetization field generated in the sample
- The demagnetization field ( $H_d = -DM$ ) act to demagnetize the magnet in opposite to the magnetization direction
- Free magnetic poles at the terminating ends of the magnet
- Strength dependent on the magnetization and the physical magnet shape



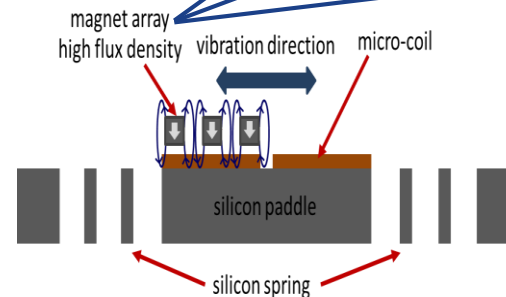
# MEMS Electromagnetic vibration energy harvester using patterned magnets

➤ **Solution:** Increasing the magnetic element edges while trying to retain the overall magnetization

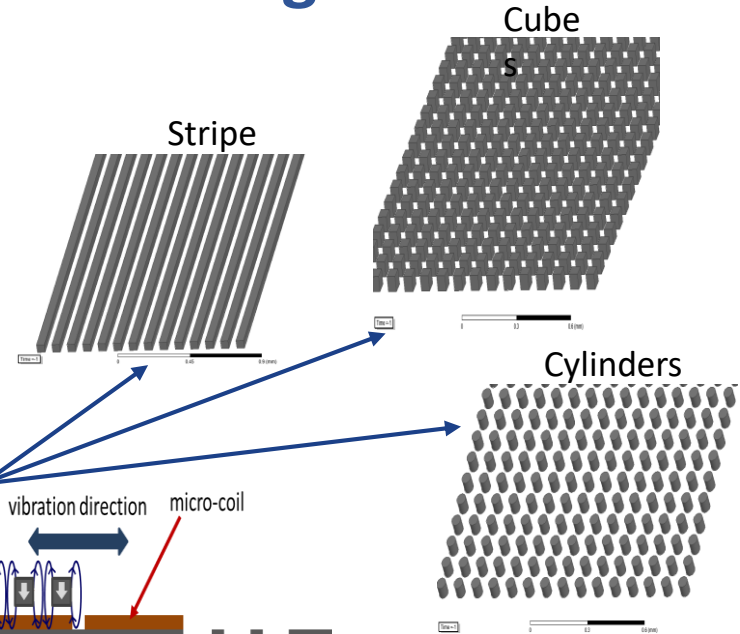
➤ Replacing thin film or block of a magnetic material by Array of Patterned Magnets



Poor performance of the integrated EM VEH due to low stray field of the block of permanent magnet



Magnetic flux density intensified over small space due to the increase of the magnetic element edges

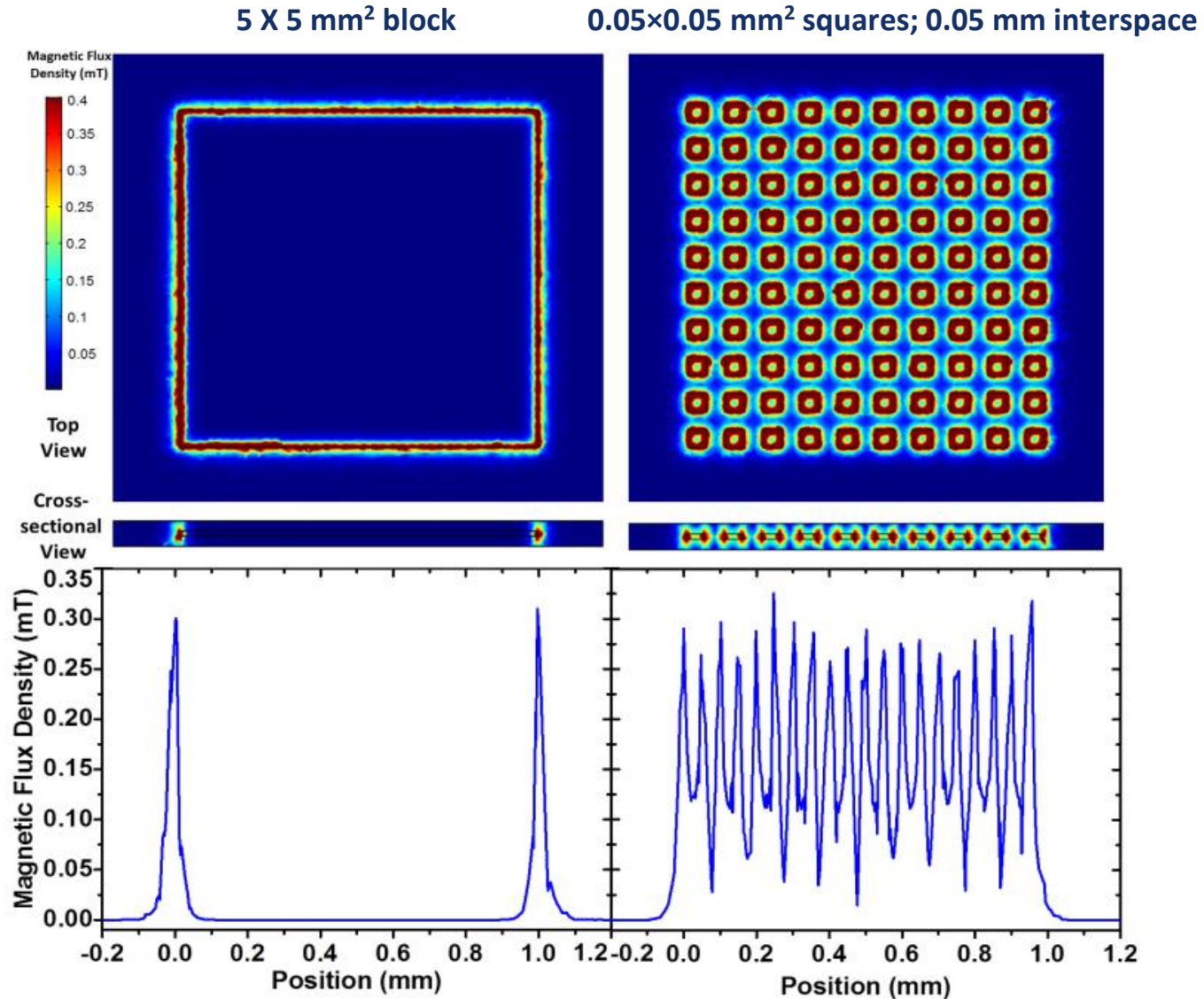


➤ **Motivation:** Development and optimization of patterned magnetic micro-structures with next-generation material that shows superior hard magnetic properties.

# Integration in MEMS EM VEH Structure

Finite Element Study on Magnetic Patterned Structures

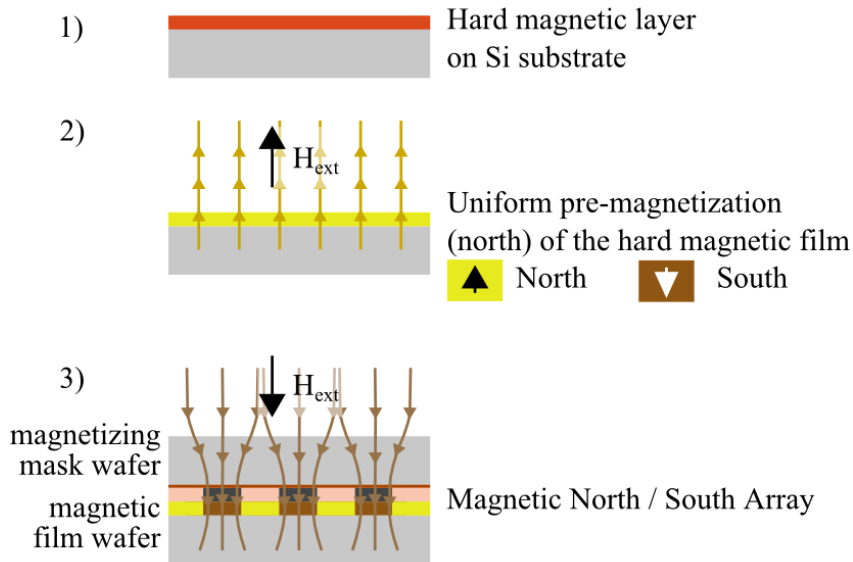
Material – CoPtP; Thickness – 50  $\mu\text{m}$



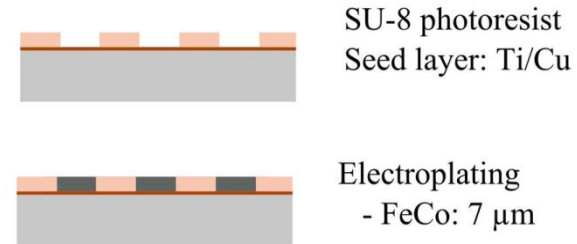
# Micro-patterning of MEMS scale PMs

## Soft magnetic mask – pulse magnetization method

### Micro-scale magnetic patterning process

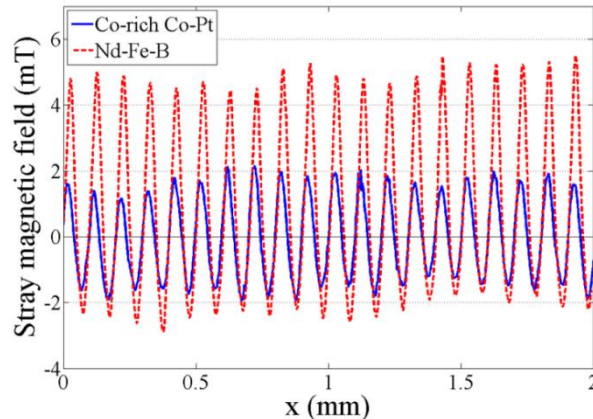


### Soft-magnetic mask fabrication



- ❖ NdFeB, CoPt PMs
- ❖ FeCo SM ( $B_{sat} = 1.3T$ )

### Magneto optical Characterization:

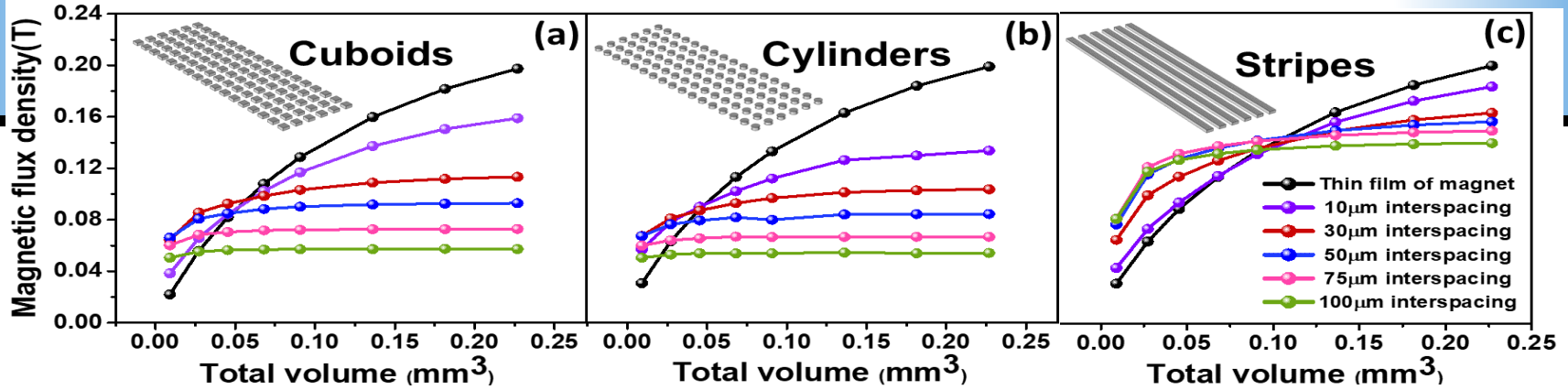


### Concerns:

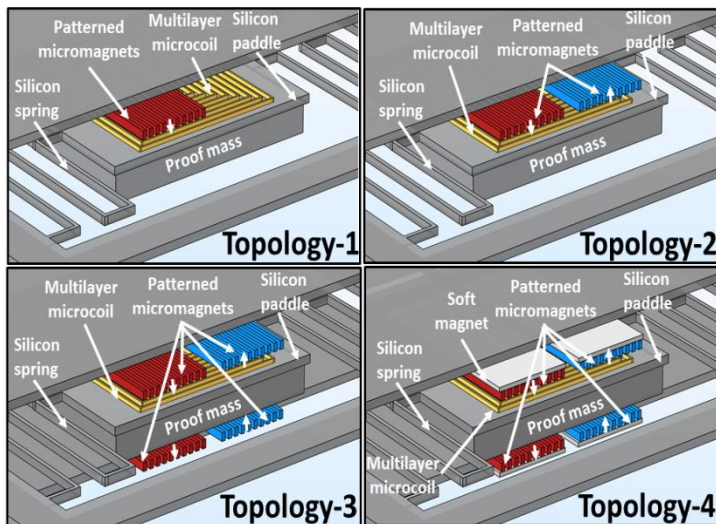
- Pulse reversal field – dependent on coercivity of PMs
- Diffraction during reverse pulse magnetization



# Optimization of the magnetic structures in terms of interspacing and aspect ratio of the array



## Four novel topologies for the MEMS EM-VEH

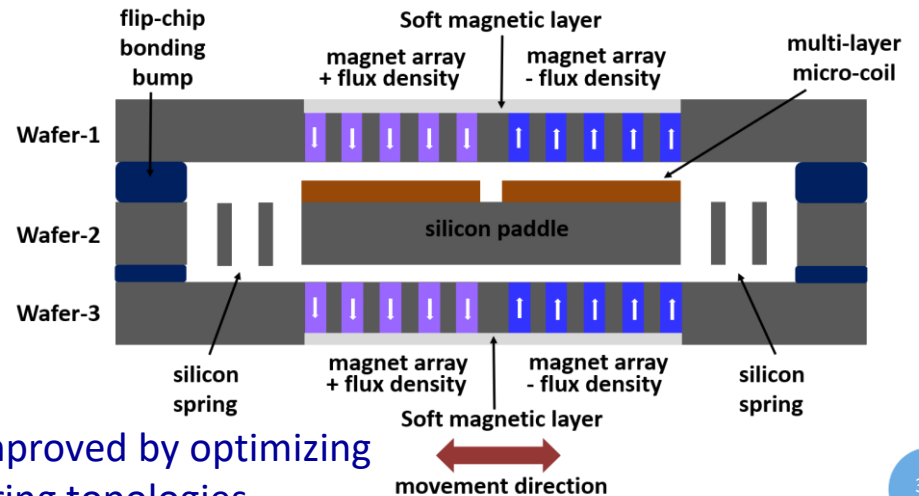
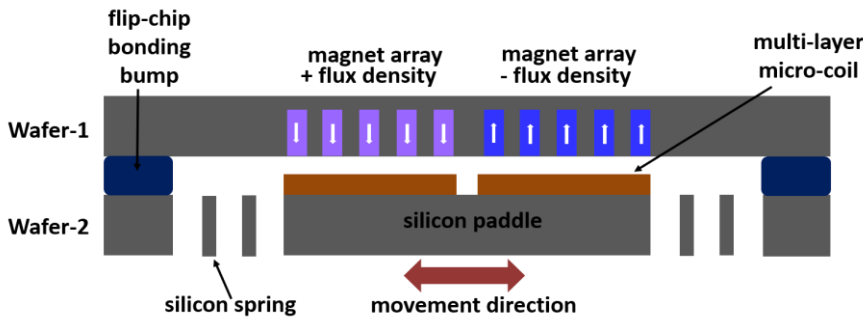
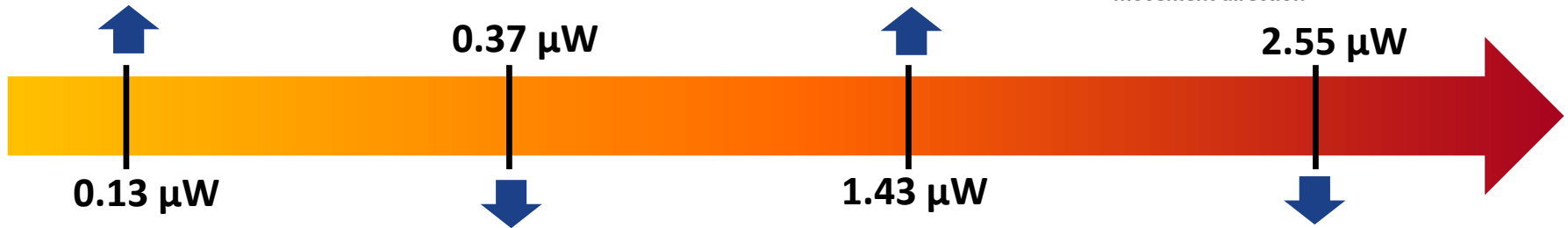
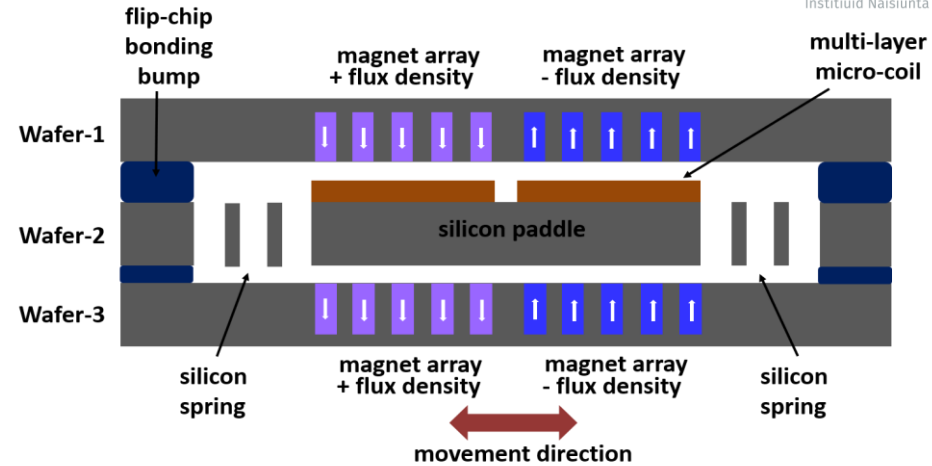
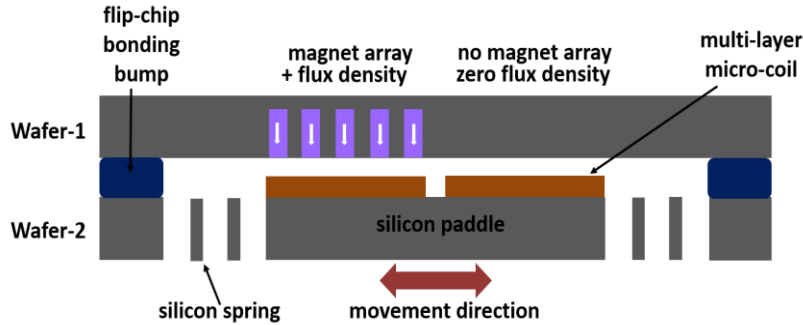


Topology	γ with Square coil (mWb/m)	γ with Rectangular coil (mWb/m)
First	2.40	10.18
Second	6.79	18.73
Third	16.21	39.40
Fourth (20 μm SM)	32.60	53.03

- ❖ IEEE Power-MEMS: K Paul, D Mallick, S Roy; Dec (2019)
- ❖ IEEE Mag Lett; K Paul, D Mallick, S Roy; 12, 1-5, (2021)

# Roadmap for Integrated EM VEH Devices

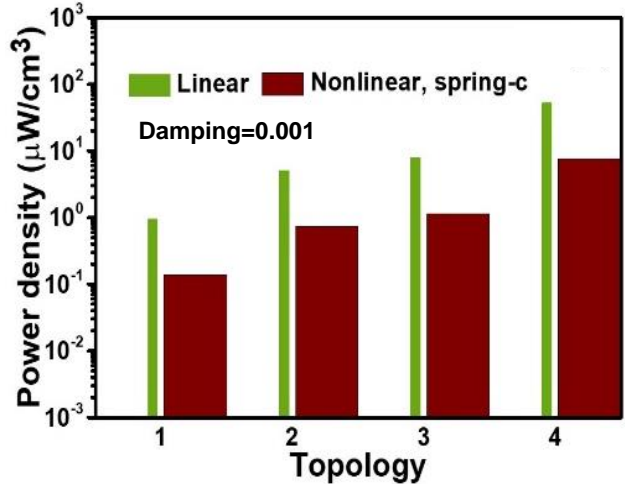
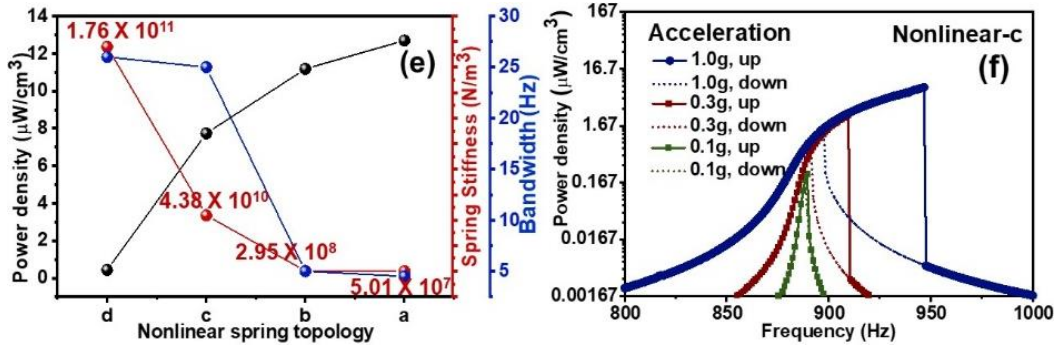
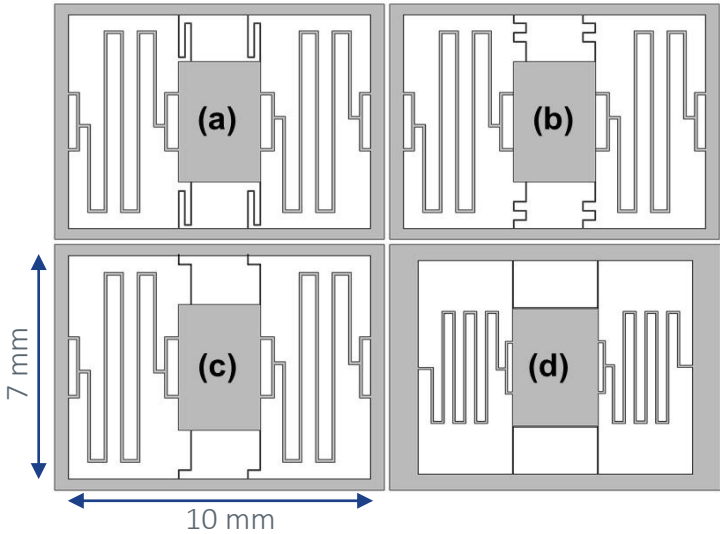
Invited Article: S Roy et al; *IEEE Trans. Mag.* 55, 4700315, (2019)



❖ Integrated devices can be further significantly improved by optimizing patterned magnets/coils and novel nonlinear spring topologies

# Performance of linear and nonlinear MEMS EM-VEH

## Comparison of different in-plane moving nonlinear spring



Reference	Volume (mm <sup>3</sup> )	Acceleration (g)	Resonant/Bandwidth	Power density
[Han 2014]	67.5	1.2	Resonant peak 48Hz	0.16μW/cm <sup>3</sup>
[Liu 2013]	35	1	Resonant peak 840Hz	0.15μW/cm <sup>3</sup>
[Liu 2014]	32	3	Resonant peak 82Hz, extended up to 146.5Hz	0.056 μW/cm <sup>3</sup>
This work, linear EM-VEH	50	1	Resonant peak 614Hz, bandwidth 4Hz	52μW/cm <sup>3</sup>
This work, Nonlinear EM-VEH (spring-c)	60	1	Non-resonant, bandwidth 25Hz (921-946Hz)	7.73μW/cm <sup>3</sup>

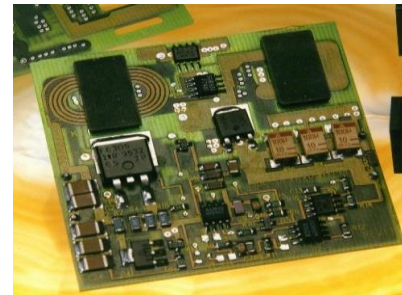
Performance comparison of state-of-the-art MEMS EM-VEHs

# **Power Conversion employing micro-nano-magnetics**

# Evolution of Integrated Power Magnetics

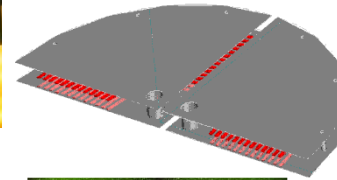


*Separate bobbins,  
windings, cores*

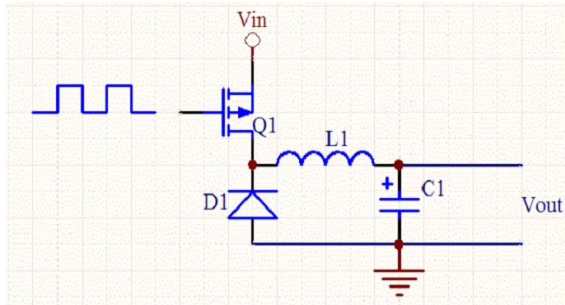


*Planar PCB Magnetics*

Integration of  
magnetics

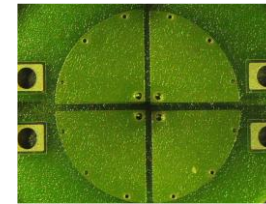


## Switch Mode Power Converter

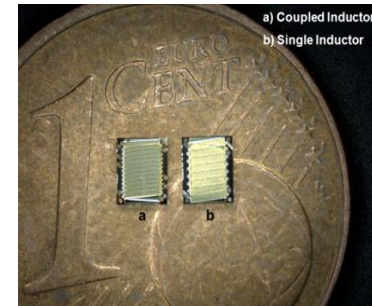


- Allowed current ripple;
- $\Delta i = (V_{in} - V_{trans} - V_{out})D / (Lf)$
- where : D = duty cycle, L= inductance, f = switching frequency.

**For small L & high Efficiency- need high f**



*PCB Integrated  
Magnetics*



*Integrated Passives  
on Si*

- **Hysteresis Loss**

- Arise from domain wall motion
- Area within  $B$ - $H$  loop

- $P_{hysteresis} = 4 \cdot f \cdot B_{ac}^2 H_c / B_{sat}$

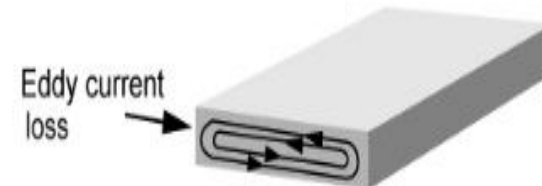
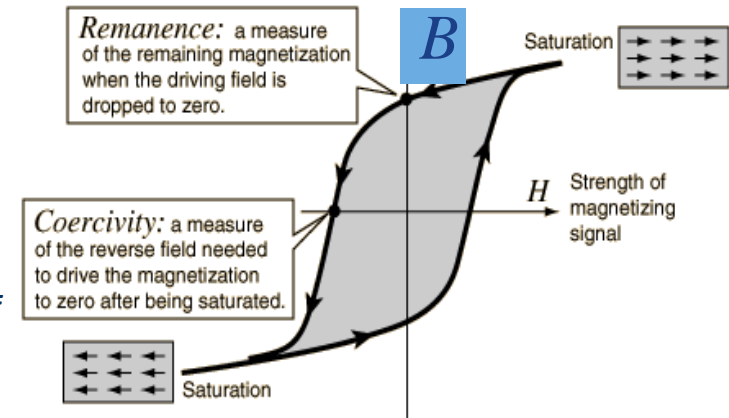
Where;  $H_c$  = coercivity,  $B$  = Magnetisation,  $H$  = Magnetising field,  $f$  = frequency

- **Eddy Current Loss**

- Eddy currents resist change in magnetic field
- Reduced by keeping thickness below 1 or 2 skin depths

- Skin depth,  $\delta = (2\rho / \omega\mu_r)^{0.5}$

Where;  $\rho$  = resistivity,  $\omega$  = angular frequency ( $=2\pi f$ )



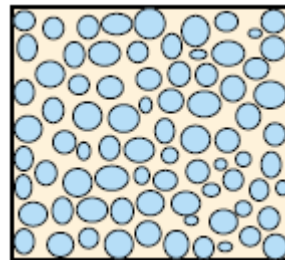
- **Anomalous Loss**

- Inconsistencies in domain wall motion

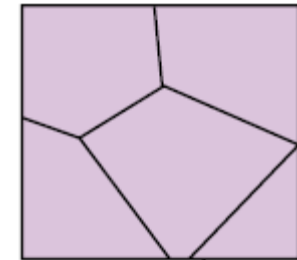
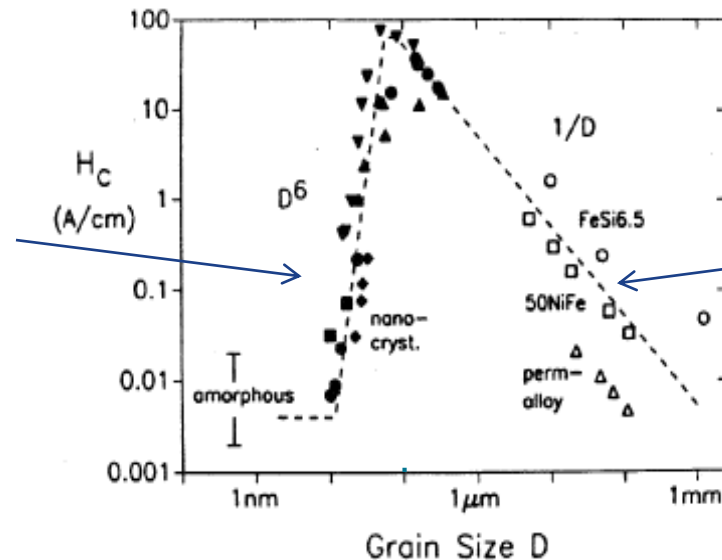
- Often calculated as;  $P_{total} = P_{hysteresis} + P_{eddy\ current} + P_{anomalous}$



# Targeted Properties for High-frequency magnetic materials



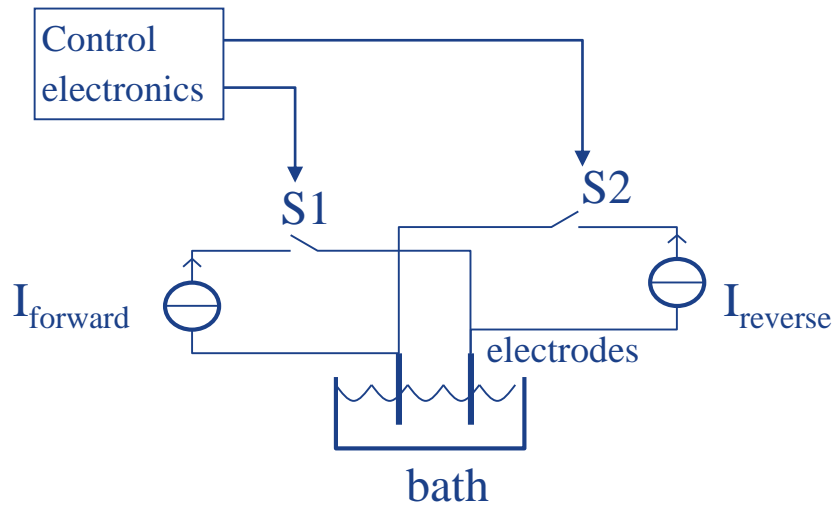
D=5 to 30nm



D > 1 $\mu$ m

- Coercivity ( $H_c$ ) < 2 Oe
- Permeability ( $\mu_r$ ) - 300-1000
- Saturation Flux density ( $B_s$ ) – 1.5- 2.4T
- Resistivity ( $\rho$ ) - 30-500  $\mu\Omega$ cm
- Cut-off frequency for eddy current loss ( $f_{ed}$ ) – 100-500 MHz
- Anisotropy field ( $H_k$ ) – 10-500e
- Natural ferromagnetic resonance frequency ( $f_{FMR}$ ) – 1-3 GHz

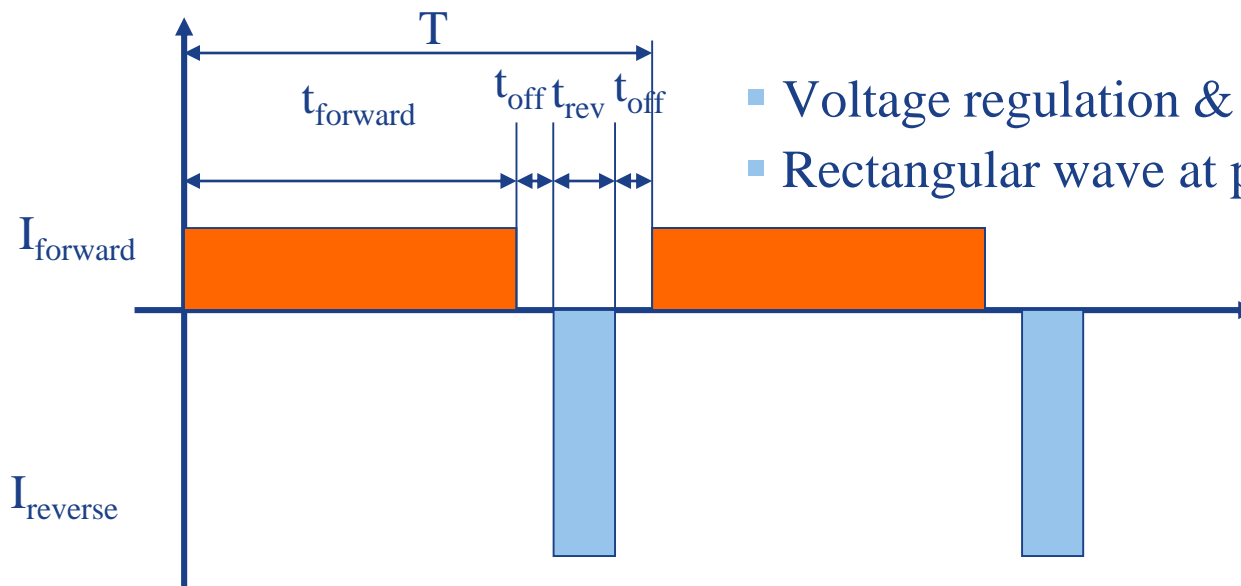
# Pulser & Pulse Reverse Plating of Magnetic Core Layers



CHI 660B Potentiostat



Wafer Plating

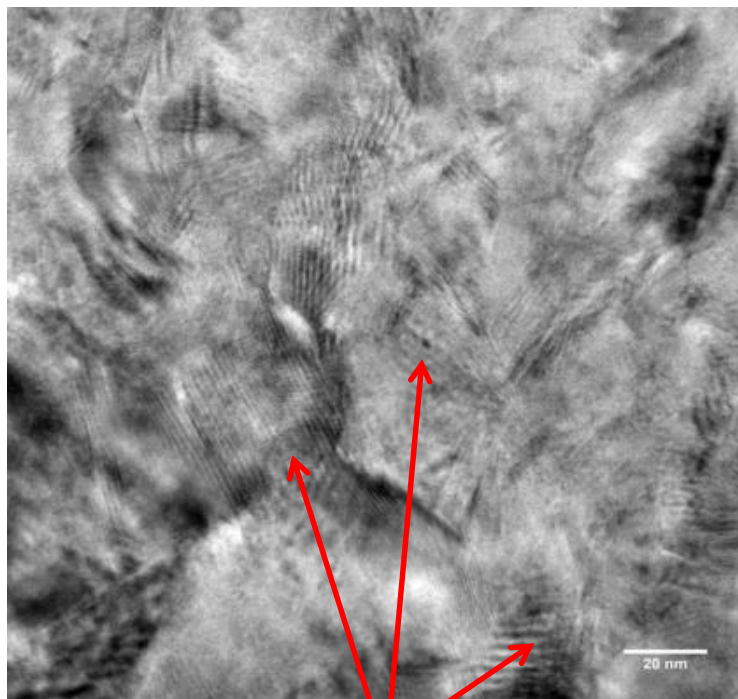


- Voltage regulation & timer circuit
- Rectangular wave at particular frequency



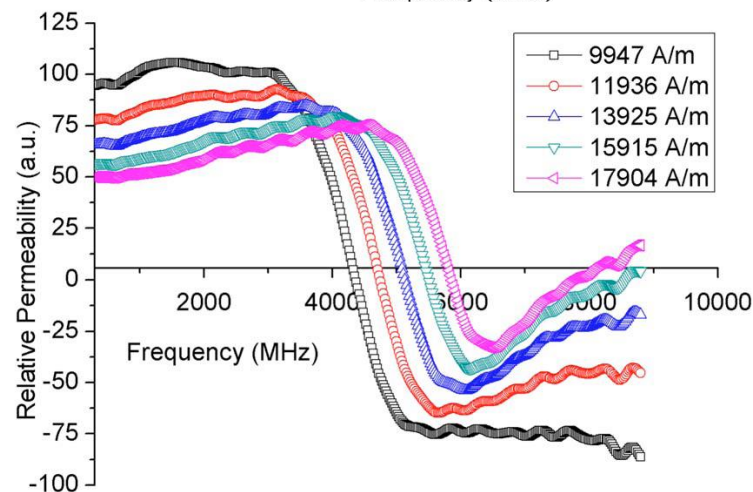
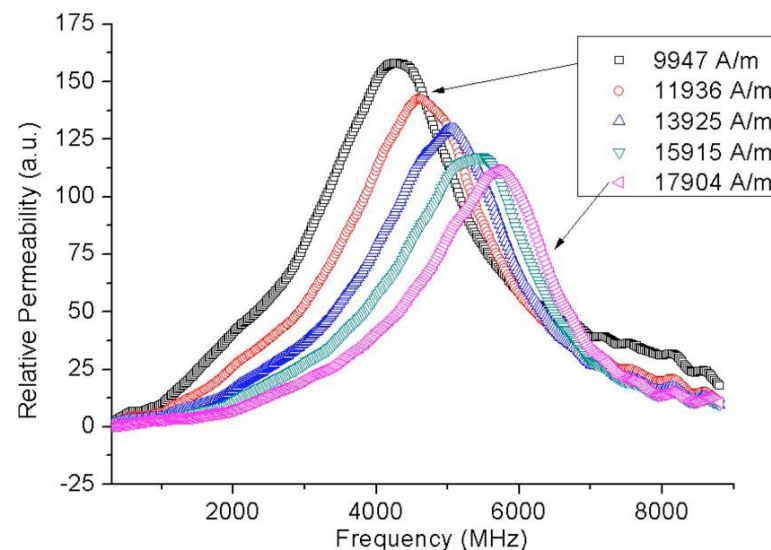
# Nanocrystalline NiFe precessional dynamics

$\text{Ni}_{45}\text{Fe}_{55}$

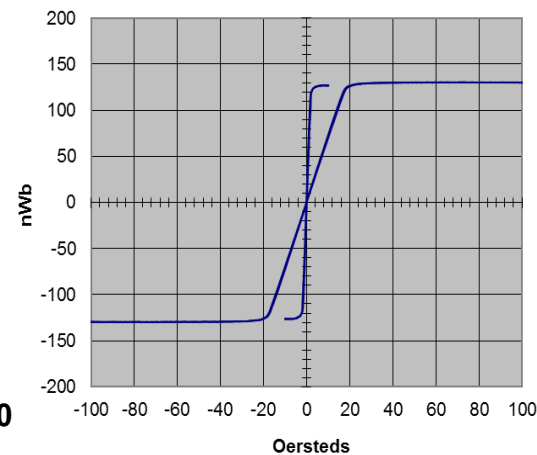
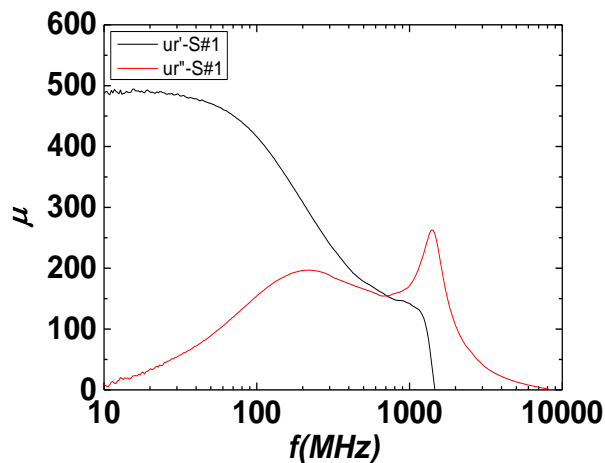
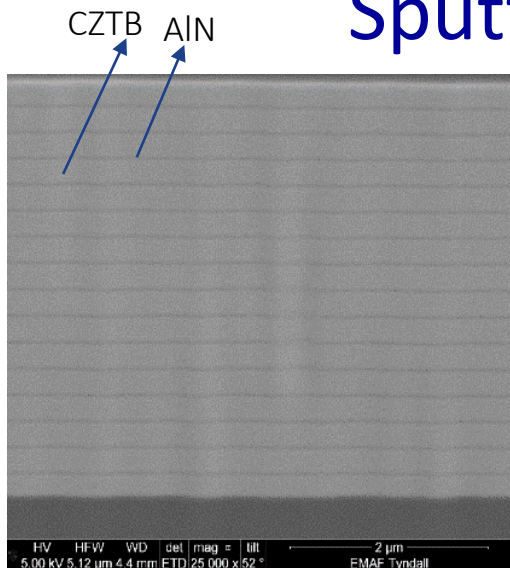


**Nanocrystalline grain structure (grains ~10nm)**

$$f_{\text{FMR}}^2 = \frac{\gamma^2 \mu_0^2 M_s}{4\pi^2} H_{\text{Bias}} + \frac{\gamma^2 \mu_0^2 M_s H_k}{4\pi^2}$$



# Sputtered CZTB Laminated Stack

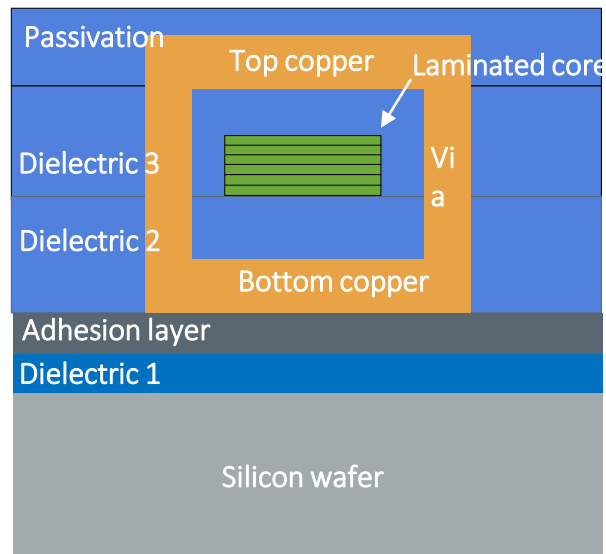


<b>Saturation, <math>B_s</math></b>	<b>1.2 T</b>
<b>Coercivity, <math>H_c</math></b>	<b>5 A/m</b>
<b>Resistivity, <math>\rho</math></b>	<b>115 <math>\mu\Omega</math> cm</b>
<b>Permeability, <math>\mu_r</math></b>	<b>490</b>

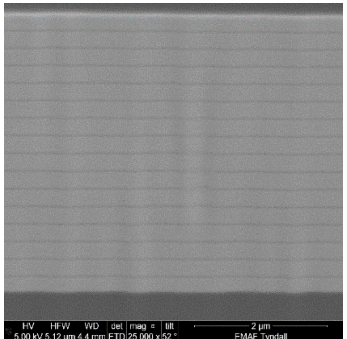
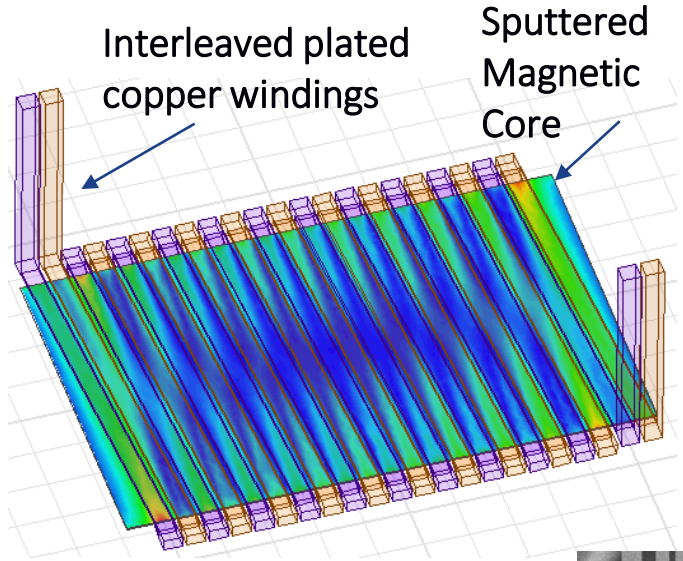
- Lower  $H_c$  => lower Hysteresis loss
- Higher resistivity => Lower eddy current loss
- Better frequency performance

❖ Lesker multi-target sputtering machine has been installed lately within Tyndall cleanroom with ~ €1M SFI Infrastructure grant to S Roy

# Solenoid Micro-Transformer with laminated core



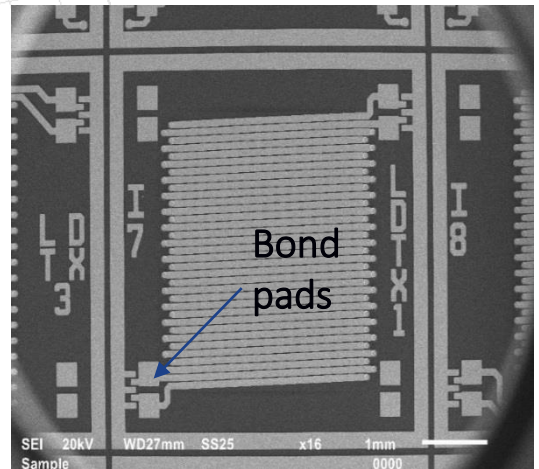
Section of solenoid structure



Laminated CZTB/AlN magnetic core

Advantages of a solenoid construction:-

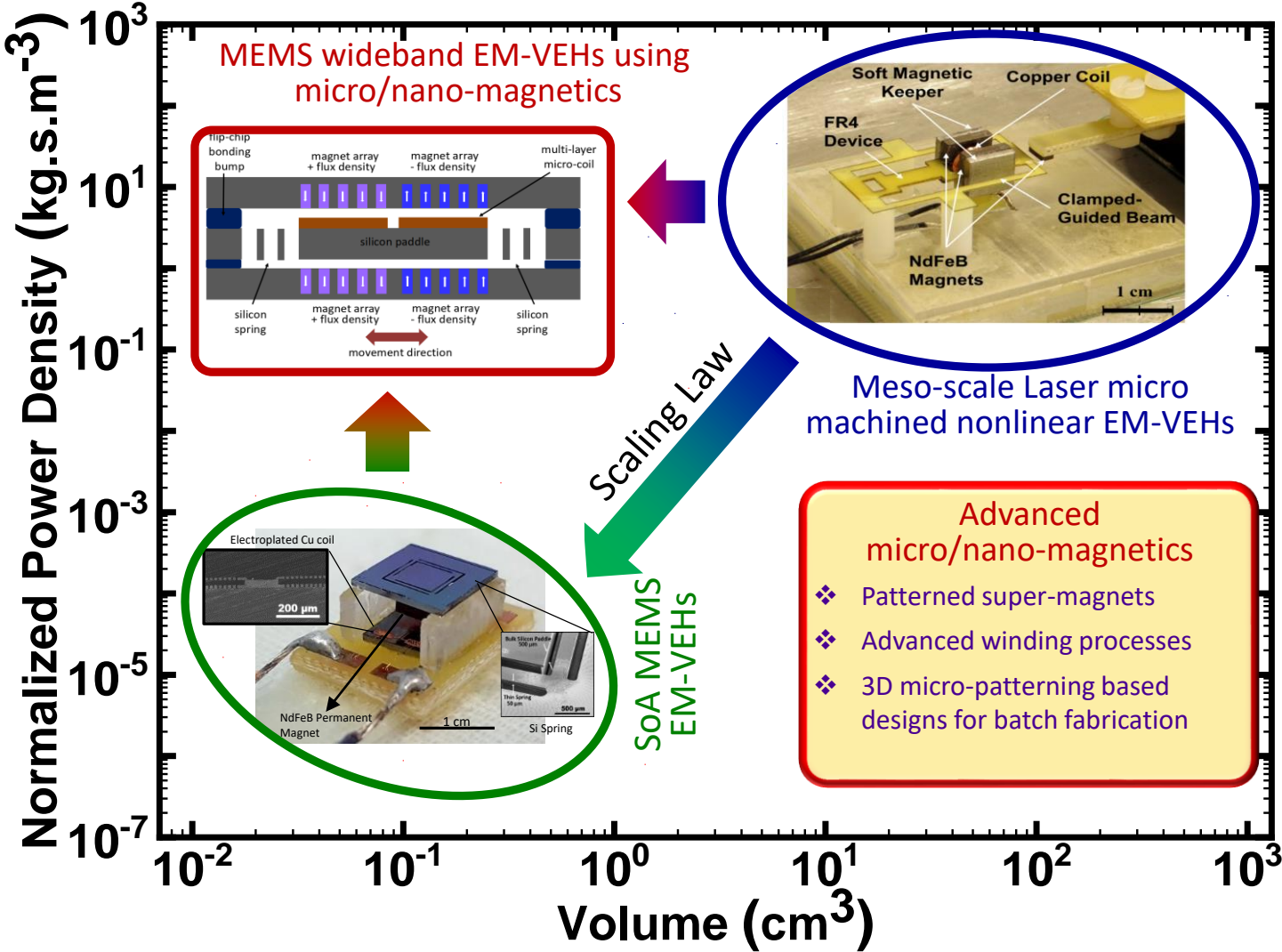
- Laminated core fabricated on a single layer (unlike a closed core which requires two layers) => Lower cost
- No magnetic via required to connect top and bottom magnetic layers



Devices prior to final passivation

# Roadmap: Future Direction

MEMS EM-VEH & conversion enabled by Micro/Nano-magnetics → Powering Internet of Things (IoT)





# Acknowledgement:

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*Ningning Wang*  
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*Dhiman Mallick*  
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*Seamus O'Driscoll*  
*Peter Constantinou*  
*Tyndall Fab Facilities*

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EU - FP6, FP7, FET Flagship Pilot  
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SFI TIDA grant award (2016) –16/TIDA/4174  
SFI Capital Infrastructure (2013) -12/RI/2342  
SFI Research Center- CONNECT- (2015 -Onwards)

## Major Collaborators:

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Prof. Terence O'Donnell- UCD, Ireland  
Prof. Steve Beeby – Univ. of Southampton, UK  
Prof. Cian O'Mathuna – Tyndall Institute, Ireland

THANK YOU